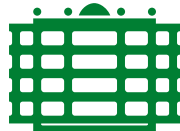


A New Test for Passivity of Descriptor Systems

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Regelungstheorie 2005

Mathematisches Forschungsinstitut Oberwolfach

27.2.–4.3.2005

Outline

- Descriptor systems
- Passive systems
- The positive real lemma
- Testing positive realness
- Conclusions and outlook



Linear Descriptor Systems

Linear time-invariant systems in generalized state-space form:

$$\begin{aligned} E\dot{x}(t) &= Ax(t) + Bu(t), \\ y(t) &= Cx(t) + Du(t), \end{aligned}$$

arise, e.g., in

- control and simulation of coupled systems,
- control of multibody (mechanical) systems,
- manipulation of fluid flow (e.g., semi-discretized Navier-Stokes equations),
- circuit simulation, VLSI chip design, in particular modeling of interconnect via RLC networks,
- simulation of MEMS and NEMS (micro-/nano-electro-mechanical systems).

Assumptions

- n generalized states / descriptor variables, i.e., $x(t) \in \mathbb{R}^n$;
- m “inputs”, i.e., $u(t) \in \mathbb{R}^m$;
- m “outputs”, i.e., $y(t) \in \mathbb{R}^m$;
- $A - \lambda E$ regular, i.e., $\exists \lambda \in \mathbb{C} : \det(A - \lambda E) \neq 0$;
- $A - \lambda E$ stable, i.e., $\lambda(A, E) \subset \mathbb{C}^- \cup \{\infty\} \Rightarrow$ system is stable.

Corresponding transfer function:

$$G(s) = C(sE - A)^{-1}B + D$$

Weierstraß Canonical Form and Index

Regular matrix pencils are equivalent $((A, E) \mapsto (PAQ, PEQ))$ to their

Weierstraß canonical form (WCF)
$$\begin{bmatrix} J & 0 \\ 0 & I_{n_\infty} \end{bmatrix} - \lambda \begin{bmatrix} I_{n_f} & 0 \\ 0 & N \end{bmatrix}.$$

Here:

- J contains Jordan blocks corresponding to the $n_f = n - n_\infty$ finite eigenvalues,
- N is nilpotent and contains Jordan blocks to the n_∞ infinite eigenvalues,
- $\nu :=$ size of largest Jordan block in N is called the (algebraic) index of $A - \lambda E$.

Fast and Slow Subsystems

Matrix pencil equivalence implies **restricted system equivalence (RSE)**:

$$(E, A, B, C, D) \sim (PEQ, PAQ, PB, CQ, D) \text{ for any nonsingular } P, Q \in \mathbb{R}^{n \times n}.$$

With the coordinate transformation $x \mapsto Px =: \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$, and a corresponding partitioning

$$PB = \begin{bmatrix} B_1 \\ B_2 \end{bmatrix}, \quad CQ = \begin{bmatrix} C_1 & C_2 \end{bmatrix},$$

RSE and the WCF imply the decoupling

$$\left. \begin{aligned} \dot{x}_1 &= Jx_1 + B_1u, \\ y_1 &= C_1x_1 + Du \end{aligned} \right\} \text{slow subsystem}$$

$$\left. \begin{aligned} N\dot{x}_2 &= x_2 + B_2u, \\ y_2 &= C_2x_2 \end{aligned} \right\} \text{fast subsystem}$$

Additive Decomposition of Transfer Function

Partial information about a descriptor system is obtained from its block-diagonal form

$$S(A - \lambda E)T = \begin{bmatrix} A_f & \\ & A_\infty \end{bmatrix} - \lambda \begin{bmatrix} E_f & \\ & E_\infty \end{bmatrix}, \quad SB = \begin{bmatrix} B_f \\ B_\infty \end{bmatrix}, \quad CT = [C_f \quad C_\infty],$$

where $S, T \in \mathbb{R}^{n \times n}$ are nonsingular. This yields

slow-fast decoupling:

$$\left. \begin{aligned} E_f \dot{x}_f &= A_f x_f + B_f u, \\ y_f &= C_f x_f + D u \end{aligned} \right\} \text{slow subsystem}$$

$$\left. \begin{aligned} E_\infty \dot{x}_\infty &= A_\infty x_\infty + B_\infty u, \\ y_\infty &= C_\infty x_\infty \end{aligned} \right\} \text{fast subsystem}$$

additive decomposition of $G(s)$:

$$G(s) = G_f(s) + G_\infty(s) = C_f (sE_f - A_f)^{-1} B_f + D + C_\infty (sE_\infty - A_\infty)^{-1} B_\infty.$$

Markov Parameters

A rational transfer function $G(s) = C(sE - A)^{-1}B + D$ has a power series expansion (Laurent series) at $s_0 = \infty$ of the form

$$G(s) = \sum_{k=-\infty}^q s^k M_k,$$

where $M_k \in \mathbb{R}^{m \times m}$ are the **Markov parameters** of G and $q \leq \nu$.

This implies

$$G(s) = G_{sp}(s) + M_0 + sM_1 + \sum_{k=2}^q s^k M_k = G_p(s) + sM_1 + \sum_{k=2}^q s^k M_k,$$

where

- $G_{sp}(s)$ is **strictly proper**, i.e., $\lim_{s \rightarrow \infty} G_{sp}(s) = 0$;
- $G_p(s) := G_{sp} + M_0$ is **proper**, i.e., $\lim_{s \rightarrow \infty} G_p(s)$ is finite.

Passive Systems

Definition:

A system is **passive** if $\int_{-\infty}^t u(\tau)^T y(\tau) d\tau \geq 0 \quad \forall t \in \mathbb{R}, \forall u \in L_2(\mathbb{R}, \mathbb{R}^m)$.

“The system cannot generate energy.”

ANDERSON/VONGPANITLERD 1973:

Theorem: system is passive \iff its transfer function is positive real

Definition:

A *real*, rational matrix-valued function $G : \mathbb{C} \rightarrow \overline{\mathbb{C}}^{m \times m}$ is (strictly) **positive real** if

1. G is analytic in $\mathbb{C}^+ := \{s \in \mathbb{C} \mid \operatorname{Re}(s) > 0\}$,
2. $G(s) + G^T(\bar{s}) \geq 0$ for all $s \in \mathbb{C}^+$ ($G(s) + G^T(\bar{s}) > 0$ for all $s \in \overline{\mathbb{C}^+}$.)

Motivation

Guaranteeing passivity of descriptor systems is necessary in

Model reduction for passive systems:

- Task often encountered in circuit simulation, VLSI chip design.
- Padé-type methods in general do not preserve passivity, post-processing necessary
[BAI/(FELDMANN)/FREUND '98,'01].
 - PRIMA [ODABASIOGLU ET AL.'96,'97] preserves passivity for interconnect models, basically Arnoldi process.
 - SyPVL preserves passivity for RLC circuits [FELDMANN/FREUND '96,'97].
 - LR-ADI/dominant subspace approximation can preserve passivity [LI/WHITE '01].
- But in general: passivity not guaranteed for reduced-order models computed by Padé-type methods, needs to be checked!

Model Reference Adaptive Control:

passivity of “input-to-tracking error” transfer function guarantees tracking property
[LANDAU 1979, DAI 1989, ...].

Hybrid Methods for Model Reduction

Recent “trend” in model reduction of LARGE-scale systems (MEMS, CFD):

“LARGE” = $n > 100,000$.

Apply method for large sparse eigenproblems like

- Jacobi-Davidson,
- Krylov-subspace method (Lanczos, Arnoldi),

to compute projector onto low-dimensional ($n \approx 1000$) subspace using

- modal truncation,
- Padé(-type) approximation,

ideas, then reduce the intermediate model further by balancing-related techniques.

For passive systems, need to check that intermediate system is still passive.

The Positive Real Lemma

KALMAN/YAKUBOVICH/POPOV/ANDERSON 1962–67

Theorem:

Let (A, B, C, D) be a minimal realization of a linear time-invariant system with transfer function $G(s)$.

a) $G(s)$ is positive real $\iff \exists$ a solution $X \geq 0$ to the LMI

$$\begin{bmatrix} A^T X + X A & X B - C^T \\ B^T X - C & -(D + D^T) \end{bmatrix} \leq 0.$$

b) If $D + D^T > 0$, then $G(s)$ is strictly positive real \iff the algebraic Riccati equation (ARE)

$$A^T X + X A + (X B - C^T)(D + D^T)^{-1}(B^T X - C) = 0,$$

has a stabilizing solution X .

A Positive Real Lemma for Descriptor Systems

FREUND/JARRE 2000/2004

Theorem:

a) **(Sufficiency)** Let (E, A, B, C, D) be a realization of a linear descriptor system with transfer function $G(s)$. If the LMIs

$$\begin{bmatrix} A^T X + X^T A & X^T B - C^T \\ B^T X - C & -(D + D^T) \end{bmatrix} \leq 0, \quad E^T X = X^T E \geq 0$$

have a solution X , then $G(s)$ is positive real.

b) **(Necessity)** Let (E, A, B, C, D) be a minimal realization of a linear descriptor system with transfer function $G(s)$ satisfying

$$D + D^T \geq M_0 + M_0^T, \quad \text{where } M_0 \text{ is the 0th Markov parameter of } G.$$

Then, if $G(s)$ is positive real, there exists a solution of the LMIs given above.

Testing Positive Realness of Descriptor Systems

- Testing positive realness via LMIs often not feasible due to computational complexity ($\mathcal{O}(n^6)$, employing structure $\mathcal{O}(n^5)$).
- Even in case $D + D^T > 0$, the Riccati equation/Hamiltonian eigenproblem test is not applicable if E is singular.
- Eigenvalue-based test for scalar transfer functions of standard and descriptor systems exists, but no generalization to MIMO is known. [BAI/FREUND 2000]
- For standard systems, recursive reduction procedure can be applied: $(A, B, C, D) \rightarrow (\tilde{A}, \tilde{B}, \tilde{C}, \tilde{D})$ with $\tilde{D} + \tilde{D}^T$ nonsingular; $G(s)$ is then positive real if $\tilde{G}(s)$ is strictly positive real. [WEISS/WANG/SPEYER 1994]

Goal: algebraic test for positive realness that only requires orthogonal coordinate transformations and has complexity $\mathcal{O}(n^3)$.

Markov Parameters and Positive Realness

ANDERSON/VONGPANITLERD 1973, FREUND/JARRE 2000:

Theorem: Given a rational matrix-valued function

$$G(s) = \underbrace{G_p(s)}_{\text{proper}} + sM_1 + \sum_{k=2}^q s^k M_k,$$

then $G(s)$ is positive real \iff $\left\{ \begin{array}{l} 1. G_p(s) \text{ is positive real,} \\ 2. M_1 \geq 0, \\ 3. M_k = 0, k = 2, 3, \dots, q. \end{array} \right.$

Remarks:

Condition 3. is trivially satisfied if the index of $A - \lambda E$ satisfies $\nu \leq 2$.

Conditions 2.+3. are trivially satisfied if $\nu \leq 1$.

Method I

Algorithm:

1. Compute an additive decomposition of $G(s)$,

$$\begin{aligned} G(s) &= G_f(s) + G_\infty(s) \\ &= [C_f(sE_f - A_f)^{-1}B_f + D] + [C_\infty(sE_\infty - A_\infty)^{-1}B_\infty]. \end{aligned}$$

2. Test if $G_\infty(s) = -C_\infty A_\infty^{-1}B_\infty - sC_\infty A_\infty^{-1}E_\infty A_\infty^{-1}B_\infty$.
3. Test if $-C_\infty A_\infty^{-1}E_\infty A_\infty^{-1}B_\infty$ is positive semidefinite.
4. Test if $G_f(s) - C_\infty A_\infty^{-1}B_\infty$ is positive real.

Remarks:

- Step 1. may be ill-conditioned as non-orthogonal transformations are required.
- Step 2. could be tested using sufficiently many interpolation points s_k so that $s_k E - A$ is well-conditioned.
- Method I is feasible for index-1 descriptor systems as Steps 2.+3. are redundant and Step 1. is “easier”.

Method II

Main ideas:

- assume minimality (otherwise, compute minimal realization first, using e.g., MATLAB Descriptor Systems Toolbox [VARGA 2000–2005]).
- avoid additive decomposition, instead compute new reduced form of $A - \lambda E$ using only orthogonal equivalence transformations;
- from reduced form obtain explicit expressions for $G_p(s)$, M_1 , and M_k for $k = 2, \dots, q$;
- the conditions $M_k = 0$ for $k \geq 2$ can be checked via the rank indices of the new reduced form;
- testing $M_1 \geq 0$ by adopted algorithm employing orthogonal decompositions only (complexity $\mathcal{O}(n^3)$);
- testing $G_p(s)$ positive real using reduction to “strictly positive real” check.

Method II: Orthogonal Reducing Equivalence Transformation

Lemma:

For any regular pencil $A - \lambda E$ there exist orthogonal matrices $U, V \in \mathbb{R}^{n \times n}$ such that

$$U(A - \lambda E)V = \begin{bmatrix} \overset{n_1}{A_{11} - \lambda E_{11}} & \overset{n_2}{A_{12} - \lambda E_{12}} & \overset{n_3}{A_{13} - \lambda E_{13}} & \overset{n_4}{A_{14} - \lambda E_{14}} \\ 0 & A_{22} & A_{23} - \lambda E_{23} & A_{24} - \lambda E_{24} \\ 0 & 0 & A_{33} & A_{34} \\ 0 & 0 & 0 & A_{44} \end{bmatrix} \begin{array}{l} \} n_1 \\ \} n_3 \\ \} n_2 \\ \} n_4 \end{array},$$

where $\text{rank}(E_{11}) = n_1$, $\text{rank}(E_{23}) = n_3$, $\text{rank}(A_{44}) = n_4$, and

$$\text{rank} \left(\begin{bmatrix} A_{22} & A_{23} - \lambda E_{23} \\ 0 & A_{33} \end{bmatrix} \right) = n_2 + n_3 \quad \forall \lambda \in \mathbb{C}.$$

Proof: constructive algorithm using 3 URVs (SVDs), 1 RQ factorization, 1 generalized Schur decomposition \rightsquigarrow complexity $\mathcal{O}(n^3)$.

Method II: Reducing Equivalence Transformation

Note: this step is only needed for theoretical purposes, no computations required!

Due to partitioning of transformed matrix pencil, as an intermediate step to WCF we can find nonsingular matrices

$$X = \begin{bmatrix} I_{n_1} & X_2 & X_3 & X_4 \\ 0 & I_{n_3} & 0 & 0 \\ 0 & 0 & I_{n_2} & 0 \\ 0 & 0 & 0 & I_{n_4} \end{bmatrix}, \quad Y = \begin{bmatrix} I_{n_1} & Y_2 & Y_3 & Y_4 \\ 0 & I_{n_2} & 0 & 0 \\ 0 & 0 & I_{n_3} & 0 \\ 0 & 0 & 0 & I_{n_4} \end{bmatrix},$$

such that

$$XU(A - \lambda E)VY = \begin{bmatrix} A_{11} - \lambda E_{11} & 0 & 0 & 0 \\ & A_{22} & A_{23} - \lambda E_{23} & A_{24} - \lambda E_{24} \\ & & A_{33} & A_{34} \\ & & & A_{44} \end{bmatrix}.$$

Method II: Representation of G_p and M_k

After the RSE transformation implied by XU, VY we get

$$G(s) = C_1(sE_{11} - A_{11})^{-1}(B_1 + X_2B_2 + X_3B_3 + X_4B_4) + D \\ + \begin{bmatrix} C_2 + C_1Y_2 & C_3 + C_1Y_3 & C_4 + C_1Y_4 \end{bmatrix} \begin{bmatrix} A_{22} & -sE_{23} + A_{23} & -sE_{24} + A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} B_2 \\ B_3 \\ B_4 \end{bmatrix}.$$

\implies formulae for G_p, M_k :

$$G_p(s) = C_1(sE_{11} - A_{11})^{-1}(B_1 + X_2B_2 + X_3B_3 + X_4B_4) + D \\ - \begin{bmatrix} C_2 + C_1Y_2 & C_3 + C_1Y_3 & C_4 + C_1Y_4 \end{bmatrix} \begin{bmatrix} A_{22} & A_{23} & A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} B_2 \\ B_3 \\ B_4 \end{bmatrix}, \\ M_k = - \begin{bmatrix} C_2 + C_1Y_2 & C_3 + C_1Y_3 & C_4 + C_1Y_4 \end{bmatrix} \left(\begin{bmatrix} A_{22} & A_{23} & A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} 0 & E_{23} & E_{24} \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \right)^k \times \\ \times \begin{bmatrix} A_{22} & A_{23} & A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} B_2 \\ B_3 \\ B_4 \end{bmatrix}, \quad k = 1, \dots, n_2 + n_3.$$

Method II: Simplification of G_p, M_k

E_{23} nonsingular \implies for $k = 1, \dots, n_2 + n_3$

$$M_k = - \begin{bmatrix} C_2 + C_1 Y_2 & C_3 + C_1 Y_3 \end{bmatrix} \left(\begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \right)^k \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} \times \\ \times \left(\begin{bmatrix} B_2 \\ B_3 \end{bmatrix} - \begin{bmatrix} A_{24} - A_{23} E_{23}^{-1} E_{24} \\ A_{34} - A_{33} E_{23}^{-1} E_{24} \end{bmatrix} A_{44}^{-1} B_4 \right).$$

With this, the transfer function becomes

$$G(s) = \hat{C} \left(s \begin{bmatrix} E_{11} & & \\ & 0 & E_{23} \\ & 0 & 0 \end{bmatrix} - \begin{bmatrix} A_{11} & & \\ & A_{22} & A_{23} \\ & 0 & A_{33} \end{bmatrix} \right)^{-1} \hat{B} + \hat{D},$$

where with notation from WCF

$$E_{11} \sim I_{n_f}, \quad A_{11} \sim J, \quad \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \sim N, \quad \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix} \sim I_{n_\infty}.$$

Method II: Testing $M_k = 0$ for $k = 2, \dots, q$

Lemma:

a) $n_2 = n_3 \implies M_k = 0$ for $k = 2, \dots, n_2 + n_3$.

b) $G(s)$ positive real $\implies n_2 = n_3$.

Proof:

$$\text{a) } n_2 = n_3 \implies \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} = \begin{bmatrix} A_{22}^{-1} & * \\ 0 & A_{33}^{-1} \end{bmatrix} \implies \left(\begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \right)^k = 0 \quad \forall k \geq 2.$$

b) [FREUND/JARRE]: G minimal, positive real \implies nilpotent matrix N from WCF

- is either empty $\implies n_2 = n_3 = 0$;
- or has only 2×2 Jordan blocks $\implies n_2 + n_3$ even.

Now, $N \sim \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \sim \begin{bmatrix} 0 & I_{n_3} \\ 0 & 0 \end{bmatrix}$, with one "1" from each Jordan block $\implies n_2 = n_3$.

Method II: Case $n_3 = 0$

$A - \lambda E$ regular $\implies n_2 = n_3 = 0 \implies M_1 = 0$ and

$$\begin{aligned} G_p(s) &= G(s) = [C_1 \quad C_4] \left(s \begin{bmatrix} E_{11} & E_{14} \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} A_{11} & A_{14} \\ 0 & A_{44} \end{bmatrix} \right)^{-1} \begin{bmatrix} B_1 \\ B_4 \end{bmatrix} + \mathcal{D}. \\ &= [C_1 \quad C_2] \left(s \begin{bmatrix} \mathcal{E}_{11} & 0 \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} \mathcal{A}_{11} & \mathcal{A}_{12} \\ \mathcal{A}_{21} & \mathcal{A}_{22} \end{bmatrix} \right)^{-1} \begin{bmatrix} \mathcal{B}_1 \\ \mathcal{B}_2 \end{bmatrix} + \mathcal{D}, \end{aligned}$$

with $\mathcal{E}_{11}, \mathcal{A}_{22}$ nonsingular.

Thus, from transformed form of $A - \lambda E$ positive realness test for $G(s)$ is reduced to testing positive realness of proper transfer function.

Method II: Case $n_2 = n_3 \neq 0$

$$\text{rank} \left(\begin{bmatrix} A_{22} & A_{23} - \lambda E_{23} \\ 0 & A_{33} \end{bmatrix} \right) = n_2 + n_3 \quad \forall \lambda \in \mathbb{C} \implies A_{22} \text{ and } A_{33} \text{ nonsingular.}$$

\implies New representation of G_p and M_1 :

$$G_p(s) = \begin{bmatrix} 0 & 0 & C_1 & C_2 & C_3 & C_4 \end{bmatrix} \left(s \begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & E_{23} & E_{24} \\ 0 & 0 & E_{11} & E_{12} & E_{13} & E_{14} \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \end{bmatrix} - \begin{bmatrix} -E_{11} & -E_{12} & 0 & 0 & 0 & 0 \\ 0 & A_{22} & 0 & 0 & 0 & 0 \\ A_{11} & A_{12} & A_{11} & A_{12} & A_{13} & A_{14} \\ 0 & 0 & 0 & A_{22} & A_{23} & A_{24} \\ 0 & 0 & 0 & 0 & A_{33} & A_{34} \\ 0 & 0 & 0 & 0 & 0 & A_{44} \end{bmatrix} \right)^{-1} \begin{bmatrix} 0 \\ 0 \\ B_1 \\ B_2 \\ B_3 \\ B_4 \end{bmatrix}$$

$$M_1 = \begin{bmatrix} C_1 & C_2 & 0 & 0 \end{bmatrix} \begin{bmatrix} E_{11} & E_{12} & 0 & 0 \\ 0 & -A_{22} & E_{23} & E_{24} \\ 0 & 0 & -A_{33} & -A_{34} \\ 0 & 0 & 0 & -A_{44} \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 0 \\ B_3 \\ B_4 \end{bmatrix}.$$

Method II: Case $n_2 = n_3 \neq 0$

$M_1 \geq 0$: using a sequence of RQ/QR/QR factorizations, we get

$$M_1 = \mathcal{N}^{-1}\mathcal{M}, \quad \mathcal{N}, \mathcal{M} \in \mathbb{R}^{m \times m}.$$

$$\implies \quad M_1 \geq 0 \iff \mathcal{M}\mathcal{N}^T \geq 0$$

$G_p(s)$ **positive real**: using a sequence of QR/RQ factorizations, we get

$$G_p(s) = [\mathcal{C}_1 \quad \mathcal{C}_2] \left(s \begin{bmatrix} \mathcal{E}_{11} & 0 \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} \mathcal{A}_{11} & \mathcal{A}_{12} \\ \mathcal{A}_{21} & \mathcal{A}_{22} \end{bmatrix} \right)^{-1} \begin{bmatrix} \mathcal{B}_1 \\ \mathcal{B}_2 \end{bmatrix} + \mathcal{D},$$

with $\mathcal{E}_{11}, \mathcal{A}_{22}$ nonsingular.

\implies Positive realness of $G_p(s)$ reduced to positive realness test for proper transfer function in standard form.

Conclusions and Outlook

- $\mathcal{O}(n^3)$ numerical algorithm for reducing the passivity test for descriptor systems to passivity test for minimal, proper transfer functions, using orthogonal RSE only.
- Based on the special form of proper systems to be tested for positive realness, we also have derived a special recursive orthogonal reduction procedure to ARE-based “strict positive realness” test.
- Implementation and thorough numerical testing necessary.
- Usage of a priori knowledge about, e.g., index or nullspace of E , can improve the performance of the algorithms.
- Method I suitable if index-1 or properness can be assumed.
- None of the approaches applicable to large, sparse descriptor systems, but sufficient for testing positive realness in model reduction methods based on modal truncation and Padé-type approximation.
- Sometimes, Writing down an LMI is not the end in computational control.