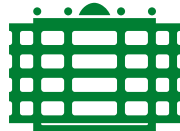


# Ein neues Kriterium zur Überprüfung der Passivität von Deskriptorsystemen

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# Outline

- Descriptor systems
- Passive systems
- The positive real lemma
- Testing positive realness
- Conclusions and outlook

## Linear Descriptor Systems

Linear time-invariant systems in generalized state-space form:

$$\begin{aligned} E\dot{x}(t) &= Ax(t) + Bu(t), \\ y(t) &= Cx(t) + Du(t), \end{aligned}$$

arise, e.g., in

- control and simulation of coupled systems,
- control of multibody (mechanical) systems,
- manipulation of fluid flow (e.g., semi-discretized Navier-Stokes equations),
- circuit simulation, VLSI chip design, in particular modeling of interconnect via RLC networks,
- simulation of MEMS and NEMS (micro-/nano-electro-mechanical systems).

## Assumptions

- $n$  generalized states / descriptor variables, i.e.,  $x(t) \in \mathbb{R}^n$ ;
- $m$  “inputs”, i.e.,  $u(t) \in \mathbb{R}^m$ ;
- $m$  “outputs”, i.e.,  $y(t) \in \mathbb{R}^m$ ;
- $A - \lambda E$  regular, i.e.,  $\exists \lambda \in \mathbb{C} : \det(A - \lambda E) \neq 0$ ;
- $A - \lambda E$  stable, i.e.,  $\lambda(A, E) \subset \mathbb{C}^- \cup \{\infty\} \Rightarrow$  system is stable.

Corresponding transfer function:

$$G(s) = C(sE - A)^{-1}B + D$$

## Weierstraß Canonical Form and Index

Regular matrix pencils are equivalent  $((A, E) \mapsto (PAQ, PEQ))$  to their

Weierstraß canonical form (WCF) 
$$\begin{bmatrix} J & 0 \\ 0 & I_{n_\infty} \end{bmatrix} - \lambda \begin{bmatrix} I_{n_f} & 0 \\ 0 & N \end{bmatrix}.$$

Here:

- $J$  contains Jordan blocks corresponding to the  $n_f = n - n_\infty$  finite eigenvalues,
- $N$  is nilpotent and contains Jordan blocks to the  $n_\infty$  infinite eigenvalues,
- $\nu :=$  size of largest Jordan block in  $N$  is called the (algebraic) index of  $A - \lambda E$ .

## Fast and Slow Subsystems

Matrix pencil equivalence implies **restricted system equivalence (RSE)**:

$$(E, A, B, C, D) \sim (PEQ, PAQ, PB, CQ, D) \text{ for any nonsingular } P, Q \in \mathbb{R}^{n \times n}.$$

With the coordinate transformation  $x \mapsto Px =: \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$ , and a corresponding partitioning

$$PB = \begin{bmatrix} B_1 \\ B_2 \end{bmatrix}, \quad CQ = \begin{bmatrix} C_1 & C_2 \end{bmatrix},$$

RSE and the WCF imply the decoupling

$$\left. \begin{aligned} \dot{x}_1 &= Jx_1 + B_1u, \\ y_1 &= C_1x_1 + Du \end{aligned} \right\} \text{ slow subsystem}$$

$$\left. \begin{aligned} N\dot{x}_2 &= x_2 + B_2u, \\ y_2 &= C_2x_2 \end{aligned} \right\} \text{ fast subsystem}$$

## Additive Decomposition of Transfer Function

Partial information about a descriptor system is obtained from its block-diagonal form

$$S(A - \lambda E)T = \begin{bmatrix} A_f & \\ & A_\infty \end{bmatrix} - \lambda \begin{bmatrix} E_f & \\ & E_\infty \end{bmatrix}, \quad SB = \begin{bmatrix} B_f \\ B_\infty \end{bmatrix}, \quad CT = [C_f \quad C_\infty],$$

where  $S, T \in \mathbb{R}^{n \times n}$  are nonsingular. This yields

**slow-fast decoupling:**

$$\begin{aligned} E_f \dot{x}_f &= A_f x_f + B_f u, \\ y_f &= C_f x_f + Du \end{aligned} \quad \left. \vphantom{\begin{aligned} E_f \dot{x}_f &= A_f x_f + B_f u, \\ y_f &= C_f x_f + Du \end{aligned}} \right\} \text{slow subsystem}$$

$$\begin{aligned} E_\infty \dot{x}_\infty &= A_\infty x_\infty + B_\infty u, \\ y_\infty &= C_\infty x_\infty \end{aligned} \quad \left. \vphantom{\begin{aligned} E_\infty \dot{x}_\infty &= A_\infty x_\infty + B_\infty u, \\ y_\infty &= C_\infty x_\infty \end{aligned}} \right\} \text{fast subsystem}$$

**additive decomposition of  $G(s)$ :**

$$G(s) = G_f(s) + G_\infty(s) = C_f(sE_f - A_f)^{-1}B_f + D + C_\infty(sE_\infty - A_\infty)^{-1}B_\infty.$$

## Markov Parameters

A rational transfer function  $G(s) = C(sE - A)^{-1}B + D$  has a power series expansion (Laurent series) at  $s_0 = \infty$  of the form

$$G(s) = \sum_{k=-\infty}^q s^k M_k,$$

where  $M_k \in \mathbb{R}^{m \times m}$  are the **Markov parameters** of  $G$  and  $q \leq \nu$ .

This implies

$$G(s) = G_{sp}(s) + M_0 + sM_1 + \sum_{k=2}^q s^k M_k = G_p(s) + sM_1 + \sum_{k=2}^q s^k M_k,$$

where

- $G_{sp}(s)$  is **strictly proper**, i.e.,  $\lim_{s \rightarrow \infty} G_{sp}(s) = 0$ ;
- $G_p(s) := G_{sp} + M_0$  is **proper**, i.e.,  $\lim_{s \rightarrow \infty} G_p(s)$  is finite.



## Passive Systems

### Definition:

A system is **passive** if  $\int_{-\infty}^t u(\tau)^T y(\tau) d\tau \geq 0 \quad \forall t \in \mathbb{R}, \forall u \in L_2(\mathbb{R}, \mathbb{R}^m)$ .

*“The system cannot generate energy.”*

ANDERSON/VONGPANITLERD 1973:

**Theorem:** system is passive  $\iff$  its transfer function is positive real

### Definition:

A *real*, rational matrix-valued function  $G : \mathbb{C} \rightarrow \overline{\mathbb{C}}^{m \times m}$  is (strictly) **positive real** if

1.  $G$  is analytic in  $\mathbb{C}^+ := \{s \in \mathbb{C} \mid \operatorname{Re}(s) > 0\}$ ,
2.  $G(s) + G^T(\bar{s}) \geq 0$  for all  $s \in \mathbb{C}^+$  ( $G(s) + G^T(\bar{s}) > 0$  for all  $s \in \overline{\mathbb{C}^+}$ .)

## Motivation

Guaranteeing passivity of descriptor systems is necessary in

### Model Reference Adaptive Control:

passivity of “input-to-tracking error” transfer function guarantees tracking property  
[LANDAU 1979, DAI 1989, ...].

**Validation** of automatically generated models of passive devices.

### Model reduction for passive systems:

- Task often encountered in circuit simulation, VLSI chip design.
- Padé-type methods in general do not preserve passivity, post-processing necessary  
[BAI/(FELDMANN)/FREUND '98,'01].
  - PRIMA [ODABASIOGLU ET AL.'96,'97] preserves passivity for interconnect models, basically Arnoldi process.
  - SyPVL preserves passivity for RLC circuits [FELDMANN/FREUND '96,'97].
  - LR-ADI/dominant subspace approximation can preserve passivity [LI/WHITE '01].
- But in general: passivity not guaranteed for reduced-order models computed by Padé-type methods!

## Hybrid Methods for Model Reduction

Recent “trend” in model reduction of LARGE-scale systems (MEMS, CFD):

“LARGE” =  $n > 100,000$ .

Apply method for large sparse eigenproblems like

- Jacobi-Davidson,
- Krylov-subspace method (Lanczos, Arnoldi),

to compute projector onto low-dimensional ( $n \approx 1000$ ) subspace using

- modal truncation,
- Padé(-type) approximation,

ideas, then reduce the intermediate model further by balancing-related techniques.

For passive systems, need to check that intermediate system is still passive.

## The Positive Real Lemma

KALMAN/YAKUBOVICH/POPOV/ANDERSON 1962–67

### Theorem:

Let  $(A, B, C, D)$  be a minimal realization of a linear time-invariant system with transfer function  $G(s)$ .

a)  $G(s)$  is positive real  $\iff \exists$  a solution  $X \geq 0$  to the LMI

$$\begin{bmatrix} A^T X + X A & X B - C^T \\ B^T X - C & -(D + D^T) \end{bmatrix} \leq 0.$$

b) If  $D + D^T > 0$ , then  $G(s)$  is strictly positive real  $\iff$  the algebraic Riccati equation (ARE)

$$A^T X + X A + (X B - C^T)(D + D^T)^{-1}(B^T X - C) = 0,$$

has a stabilizing solution  $X$ .

## A Positive Real Lemma for Descriptor Systems

FREUND/JARRE 2000/2004

### Theorem:

a) **(Sufficiency)** Let  $(E, A, B, C, D)$  be a realization of a linear descriptor system with transfer function  $G(s)$ . If the LMIs

$$\begin{bmatrix} A^T X + X^T A & X^T B - C^T \\ B^T X - C & -(D + D^T) \end{bmatrix} \leq 0, \quad E^T X = X^T E \geq 0$$

have a solution  $X$ , then  $G(s)$  is positive real.

b) **(Necessity)** Let  $(E, A, B, C, D)$  be a minimal realization of a linear descriptor system with transfer function  $G(s)$  satisfying

$$D + D^T \geq M_0 + M_0^T, \quad \text{where } M_0 \text{ is the 0th Markov parameter of } G.$$

Then, if  $G(s)$  is positive real, there exists a solution of the LMIs given above.

## Testing Positive Realness of Descriptor Systems

- Testing positive realness via LMIs often not feasible due to computational complexity ( $\mathcal{O}(n^6)$ , employing structure  $\mathcal{O}(n^5)$ ).
- Even in case  $D + D^T > 0$ , the Riccati equation/Hamiltonian eigenproblem test is not applicable if  $E$  is singular.
- Eigenvalue-based test for scalar transfer functions of standard and descriptor systems exists, but no generalization to MIMO is known. [BAI/FREUND 2000]
- For standard systems, recursive reduction procedure can be applied:  $(A, B, C, D) \rightarrow (\tilde{A}, \tilde{B}, \tilde{C}, \tilde{D})$  with  $\tilde{D} + \tilde{D}^T$  nonsingular;  $G(s)$  is then positive real if  $\tilde{G}(s)$  is strictly positive real. [WEISS/WANG/SPEYER 1994]

**Goal:** algebraic test for positive realness that only requires orthogonal coordinate transformations and has complexity  $\mathcal{O}(n^3)$ .

## Markov Parameters and Positive Realness

ANDERSON/VONGPANITLERD 1973, FREUND/JARRE 2000:

**Theorem:** Given a rational matrix-valued function

$$G(s) = \underbrace{G_p(s)}_{\text{proper}} + sM_1 + \sum_{k=2}^q s^k M_k,$$

then  $G(s)$  is positive real  $\iff$   $\left\{ \begin{array}{l} 1. G_p(s) \text{ is positive real,} \\ 2. M_1 \geq 0, \\ 3. M_k = 0, k = 2, 3, \dots, q. \end{array} \right.$

### Remarks:

Condition 3. is trivially satisfied if the index of  $A - \lambda E$  satisfies  $\nu \leq 2$ .

Conditions 2.+3. are trivially satisfied if  $\nu \leq 1$ .

## Method I

### Algorithm:

1. Compute an additive decomposition of  $G(s)$ ,

$$\begin{aligned} G(s) &= G_f(s) + G_\infty(s) \\ &= [C_f(sE_f - A_f)^{-1}B_f + D] + [C_\infty(sE_\infty - A_\infty)^{-1}B_\infty]. \end{aligned}$$

2. Test if  $G_\infty(s) = -C_\infty A_\infty^{-1}B_\infty - sC_\infty A_\infty^{-1}E_\infty A_\infty^{-1}B_\infty$ .
3. Test if  $-C_\infty A_\infty^{-1}E_\infty A_\infty^{-1}B_\infty$  is positive semidefinite.
4. Test if  $G_f(s) - C_\infty A_\infty^{-1}B_\infty$  is positive real.

### Remarks:

- Step 1. may be ill-conditioned as non-orthogonal transformations are required.
- Step 2. could be tested using sufficiently many interpolation points  $s_k$  so that  $s_k E - A$  is well-conditioned.
- Method I is feasible for index-1 descriptor systems as Steps 2.+3. are redundant and Step 1. is “easier”.



## Method II

### Main ideas:

- assume minimality (otherwise, compute minimal realization first, using e.g., MATLAB Descriptor Systems Toolbox [VARGA 1999–2005]).
- avoid additive decomposition, instead compute new reduced form of  $A - \lambda E$  using only orthogonal equivalence transformations;
- from reduced form obtain explicit expressions for  $G_p(s)$ ,  $M_1$ , and  $M_k$  for  $k = 2, \dots, q$ ;
- the conditions  $M_k = 0$  for  $k \geq 2$  can be checked via the rank indices of the new reduced form;
- testing  $M_1 \geq 0$  by adopted algorithm employing orthogonal decompositions only (complexity  $\mathcal{O}(n^3)$ );
- testing  $G_p(s)$  positive real using reduction to “strictly positive real” check.

## Method II: Orthogonal Reducing Equivalence Transformation

### Lemma:

For any regular pencil  $A - \lambda E$  there exist orthogonal matrices  $U, V \in \mathbb{R}^{n \times n}$  such that

$$U(A - \lambda E)V = \begin{bmatrix} \overset{n_1}{A_{11} - \lambda E_{11}} & \overset{n_2}{A_{12} - \lambda E_{12}} & \overset{n_3}{A_{13} - \lambda E_{13}} & \overset{n_4}{A_{14} - \lambda E_{14}} \\ 0 & A_{22} & A_{23} - \lambda E_{23} & A_{24} - \lambda E_{24} \\ 0 & 0 & A_{33} & A_{34} \\ 0 & 0 & 0 & A_{44} \end{bmatrix} \begin{matrix} \} n_1 \\ \} n_3 \\ \} n_2 \\ \} n_4 \end{matrix},$$

where  $\text{rank}(E_{11}) = n_1$ ,  $\text{rank}(E_{23}) = n_3$ ,  $\text{rank}(A_{44}) = n_4$ , and

$$\text{rank} \left( \begin{bmatrix} A_{22} & A_{23} - \lambda E_{23} \\ 0 & A_{33} \end{bmatrix} \right) = n_2 + n_3 \quad \forall \lambda \in \mathbb{C}.$$

**Proof:** constructive algorithm using 3 URVs (SVDs), 1 RQ factorization, 1 generalized Schur decomposition  $\rightsquigarrow$  complexity  $\mathcal{O}(n^3)$ .

## Method II: Reducing Equivalence Transformation

Note: this step is only needed for theoretical purposes, no computations required!

Due to partitioning of transformed matrix pencil, as an intermediate step to WCF we can find nonsingular matrices

$$X = \begin{bmatrix} I_{n_1} & X_2 & X_3 & X_4 \\ 0 & I_{n_3} & 0 & 0 \\ 0 & 0 & I_{n_2} & 0 \\ 0 & 0 & 0 & I_{n_4} \end{bmatrix}, \quad Y = \begin{bmatrix} I_{n_1} & Y_2 & Y_3 & Y_4 \\ 0 & I_{n_2} & 0 & 0 \\ 0 & 0 & I_{n_3} & 0 \\ 0 & 0 & 0 & I_{n_4} \end{bmatrix},$$

such that

$$XU(A - \lambda E)VY = \begin{bmatrix} A_{11} - \lambda E_{11} & 0 & 0 & 0 \\ & A_{22} & A_{23} - \lambda E_{23} & A_{24} - \lambda E_{24} \\ & & A_{33} & A_{34} \\ & & & A_{44} \end{bmatrix}.$$

## Method II: Representation of $G_p$ and $M_k$

After the RSE transformation implied by  $XU, VY$  we get

$$G(s) = C_1(sE_{11} - A_{11})^{-1}(B_1 + X_2B_2 + X_3B_3 + X_4B_4) + D \\ + \begin{bmatrix} C_2 + C_1Y_2 & C_3 + C_1Y_3 & C_4 + C_1Y_4 \end{bmatrix} \begin{bmatrix} A_{22} & -sE_{23} + A_{23} & -sE_{24} + A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} B_2 \\ B_3 \\ B_4 \end{bmatrix}.$$

$\implies$  formulae for  $G_p, M_k$ :

$$G_p(s) = C_1(sE_{11} - A_{11})^{-1}(B_1 + X_2B_2 + X_3B_3 + X_4B_4) + D \\ - \begin{bmatrix} C_2 + C_1Y_2 & C_3 + C_1Y_3 & C_4 + C_1Y_4 \end{bmatrix} \begin{bmatrix} A_{22} & A_{23} & A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} B_2 \\ B_3 \\ B_4 \end{bmatrix}, \\ M_k = - \begin{bmatrix} C_2 + C_1Y_2 & C_3 + C_1Y_3 & C_4 + C_1Y_4 \end{bmatrix} \left( \begin{bmatrix} A_{22} & A_{23} & A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} 0 & E_{23} & E_{24} \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \right)^k \times \\ \times \begin{bmatrix} A_{22} & A_{23} & A_{24} \\ 0 & A_{33} & A_{34} \\ 0 & 0 & A_{44} \end{bmatrix}^{-1} \begin{bmatrix} B_2 \\ B_3 \\ B_4 \end{bmatrix}, \quad k = 1, \dots, n_2 + n_3.$$

## Method II: Simplification of $G_p, M_k$

$E_{23}$  nonsingular  $\implies$  for  $k = 1, \dots, n_2 + n_3$

$$M_k = - \begin{bmatrix} C_2 + C_1 Y_2 & C_3 + C_1 Y_3 \end{bmatrix} \left( \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \right)^k \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} \times \\ \times \left( \begin{bmatrix} B_2 \\ B_3 \end{bmatrix} - \begin{bmatrix} A_{24} - A_{23} E_{23}^{-1} E_{24} \\ A_{34} - A_{33} E_{23}^{-1} E_{24} \end{bmatrix} A_{44}^{-1} B_4 \right).$$

With this, the transfer function becomes

$$G(s) = \hat{C} \left( s \begin{bmatrix} E_{11} & & \\ & 0 & E_{23} \\ & 0 & 0 \end{bmatrix} - \begin{bmatrix} A_{11} & & \\ & A_{22} & A_{23} \\ & 0 & A_{33} \end{bmatrix} \right)^{-1} \hat{B} + \hat{D},$$

where with notation from WCF

$$E_{11} \sim I_{n_f}, \quad A_{11} \sim J, \quad \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \sim N, \quad \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix} \sim I_{n_\infty}.$$

## Method II: Testing $M_k = 0$ for $k = 2, \dots, q$

### Lemma:

a)  $n_2 = n_3 \implies M_k = 0$  for  $k = 2, \dots, n_2 + n_3$ .

b)  $G(s)$  positive real  $\implies n_2 = n_3$ .

### Proof:

a)  $n_2 = n_3 \implies \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} = \begin{bmatrix} A_{22}^{-1} & * \\ 0 & A_{33}^{-1} \end{bmatrix} \implies \left( \begin{bmatrix} A_{22} & A_{23} \\ 0 & A_{33} \end{bmatrix}^{-1} \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \right)^k = 0 \forall k \geq 2.$

b) [FREUND/JARRE]:  $G$  minimal, positive real  $\implies$  nilpotent matrix  $N$  from WCF

- is either empty  $\implies n_2 = n_3 = 0$ ;
- or has only  $2 \times 2$  Jordan blocks  $\implies n_2 + n_3$  even.

Now,  $N \sim \begin{bmatrix} 0 & E_{23} \\ 0 & 0 \end{bmatrix} \sim \begin{bmatrix} 0 & I_{n_3} \\ 0 & 0 \end{bmatrix}$ , with one “1” from each Jordan block  $\implies n_2 = n_3$ .

## Method II: Case $n_3 = 0$

$A - \lambda E$  regular  $\implies n_2 = n_3 = 0 \implies M_1 = 0$  and

$$\begin{aligned} G_p(s) &= G(s) = [C_1 \quad C_4] \left( s \begin{bmatrix} E_{11} & E_{14} \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} A_{11} & A_{14} \\ 0 & A_{44} \end{bmatrix} \right)^{-1} \begin{bmatrix} B_1 \\ B_4 \end{bmatrix} + \mathcal{D}. \\ &= [C_1 \quad C_2] \left( s \begin{bmatrix} \mathcal{E}_{11} & 0 \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} \mathcal{A}_{11} & \mathcal{A}_{12} \\ \mathcal{A}_{21} & \mathcal{A}_{22} \end{bmatrix} \right)^{-1} \begin{bmatrix} \mathcal{B}_1 \\ \mathcal{B}_2 \end{bmatrix} + \mathcal{D}, \end{aligned}$$

with  $\mathcal{E}_{11}, \mathcal{A}_{22}$  nonsingular.

Thus, from transformed form of  $A - \lambda E$  positive realness test for  $G(s)$  is reduced to testing positive realness of proper transfer function.

## Method II: Case $n_2 = n_3 \neq 0$

$$\text{rank} \left( \begin{bmatrix} A_{22} & A_{23} - \lambda E_{23} \\ 0 & A_{33} \end{bmatrix} \right) = n_2 + n_3 \quad \forall \lambda \in \mathbb{C} \implies A_{22} \text{ and } A_{33} \text{ nonsingular.}$$

$\implies$  New representation of  $G_p$  and  $M_1$ :

$$G_p(s) = \begin{bmatrix} 0 & 0 & C_1 & C_2 & C_3 & C_4 \end{bmatrix} \left( s \begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & E_{23} & E_{24} \\ 0 & 0 & E_{11} & E_{12} & E_{13} & E_{14} \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \end{bmatrix} - \begin{bmatrix} -E_{11} & -E_{12} & 0 & 0 & 0 & 0 \\ 0 & A_{22} & 0 & 0 & 0 & 0 \\ A_{11} & A_{12} & A_{11} & A_{12} & A_{13} & A_{14} \\ 0 & 0 & 0 & A_{22} & A_{23} & A_{24} \\ 0 & 0 & 0 & 0 & A_{33} & A_{34} \\ 0 & 0 & 0 & 0 & 0 & A_{44} \end{bmatrix} \right)^{-1} \begin{bmatrix} 0 \\ 0 \\ B_1 \\ B_2 \\ B_3 \\ B_4 \end{bmatrix}$$

$$M_1 = \begin{bmatrix} C_1 & C_2 & 0 & 0 \end{bmatrix} \begin{bmatrix} E_{11} & E_{12} & 0 & 0 \\ 0 & -A_{22} & E_{23} & E_{24} \\ 0 & 0 & -A_{33} & -A_{34} \\ 0 & 0 & 0 & -A_{44} \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 0 \\ B_3 \\ B_4 \end{bmatrix}.$$



## Method II: Case $n_2 = n_3 \neq 0$

$M_1 \geq 0$ : using a sequence of RQ/QR/QR factorizations, we get

$$M_1 = \mathcal{N}^{-1}\mathcal{M}, \quad \mathcal{N}, \mathcal{M} \in \mathbb{R}^{m \times m}.$$

$$\implies \quad M_1 \geq 0 \iff \mathcal{M}\mathcal{N}^T \geq 0$$

$G_p(s)$  **positive real**: using a sequence of QR/RQ factorizations, we get

$$G_p(s) = [ \mathcal{C}_1 \quad \mathcal{C}_2 ] \left( s \begin{bmatrix} \mathcal{E}_{11} & 0 \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} \mathcal{A}_{11} & \mathcal{A}_{12} \\ \mathcal{A}_{21} & \mathcal{A}_{22} \end{bmatrix} \right)^{-1} \begin{bmatrix} \mathcal{B}_1 \\ \mathcal{B}_2 \end{bmatrix} + \mathcal{D},$$

with  $\mathcal{E}_{11}, \mathcal{A}_{22}$  nonsingular.

$\implies$  Positive realness of  $G_p(s)$  reduced to positive realness test for proper transfer function in standard form.

## Conclusions and Outlook

- $\mathcal{O}(n^3)$  numerical algorithm for reducing the passivity test for descriptor systems to passivity test for minimal, proper transfer functions, using orthogonal RSE only.
- Based on the special form of proper systems to be tested for positive realness, we also have derived a special recursive orthogonal reduction procedure to ARE-based “strict positive realness” test.
- Implementation and thorough numerical testing necessary.
- Usage of a priori knowledge about, e.g., index or nullspace of  $E$ , can improve the performance of the algorithms.
- Method I suitable if index-1 or properness can be assumed.
- None of the approaches applicable to large, sparse descriptor systems, but sufficient for testing positive realness in model reduction methods based on modal truncation and Padé-type approximation.
- Sometimes, writing down an LMI is not the end in computational control.