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SYSTEM-THEORETIC MODEL REDUCTION FOR NONLINEAR SYSTEMS

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Overview



Introduction

- Model Reduction for Dynamical Systems
- Application Areas
- Motivating Examples
- Nonlinear Model Reduction
- 2 \mathcal{H}_2 -Model Reduction for Bilinear Systems
- 8 Nonlinear Model Reduction by Generalized Moment-Matching
- 4 Numerical Examples
- 5 Conclusions and Outlook

Model Reduction for Dynamical Systems



Dynamical Systems

$$\Sigma : \begin{cases} \dot{x}(t) = f(t, x(t), u(t)), & x(t_0) = x_0, \\ y(t) = g(t, x(t), u(t)) \end{cases}$$

with

- states $x(t) \in \mathbb{R}^n$,
- inputs $u(t) \in \mathbb{R}^m$,
- outputs $y(t) \in \mathbb{R}^p$.



Model Reduction for Dynamical Systems



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Goal:

 $||y - \hat{y}|| < \text{tolerance} \cdot ||u||$ for all admissible input signals.

\mathcal{H}_2 -Model Reduction for Bilinear Systems Nonlinear Model Reduction Numerical Examples Conclusions and Outlook References

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Reduced-Order Model (ROM)

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- states $\hat{x}(t) \in \mathbb{R}^r$, $r \ll n$
- inputs $u(t) \in \mathbb{R}^m$,





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u

<u>u</u> <u>y</u>

Goal:

 $||y - \hat{y}|| < \text{tolerance} \cdot ||u||$ for all admissible input signals. Secondary goal: reconstruct approximation of x from \hat{x} .

Model Reduction for Dynamical Systems

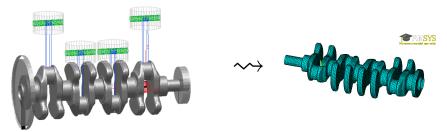


Linear, Time-Invariant (LTI) Systems

$$\begin{array}{rcl} E\dot{x} &=& f(t,x,u) &=& Ax + Bu, \quad E, A \in \mathbb{R}^{n \times n}, \\ y &=& g(t,x,u) &=& Cx + Du, \quad C \in \mathbb{R}^{p \times n}, \end{array} \qquad \begin{array}{rcl} B \in \mathbb{R}^{n \times m} \\ D \in \mathbb{R}^{p \times m} \end{array}$$

Application Areas Structural Mechanics / Finite Element Modeling

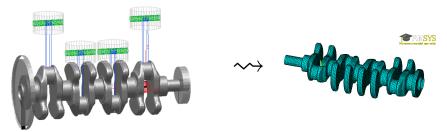




- Resolving complex 3D geometries \Rightarrow millions of degrees of freedom.
- Analysis of elastic deformations requires many simulation runs for varying external forces.

Application Areas Structural Mechanics / Finite Element Modeling





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Standard MOR techniques in structural mechanics: modal truncation, combined with Guyan reduction (static condensation) \rightsquigarrow Craig-Bampton method.

Application Areas (Optimal) Control

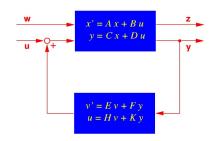


Feedback Controllers

A feedback controller (dynamic compensator) is a linear system of order N, where

- input = output of plant,
- output = input of plant.

Modern (LQG- $/\mathcal{H}_2$ - $/\mathcal{H}_\infty$ -) control design: $N \ge n$.



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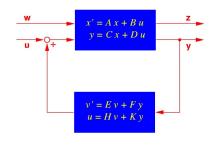


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Practical controllers require small N (N \sim 10, say) due to

- real-time constraints,
- increasing fragility for larger N.

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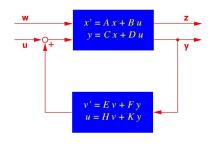


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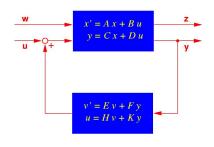


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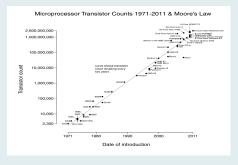
Standard MOR techniques in systems and control: balanced truncation and related methods.

Micro Electronics/Circuit Simulation



Progressive miniaturization

- Verification of VLSI/ULSI chip design requires high number of simulations for different input signals.
- Moore's Law (1965/75) states that the number of on-chip transistors doubles each 24 months.



Source: http://en.wikipedia.org/wiki/File:Transistor_Count_and_Moore'sLaw_-_2011.svg

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- Increase in packing density and multilayer technology requires modeling of interconncet to ensure that thermic/electro-magnetic effects do not disturb signal transmission.

Intel 4004 (1971)	Intel Core 2 Extreme (quad-core) (2007)
1 layer, 10 μ technology	9 layers, 45 <i>nm</i> technology
	> 8, 200, 000 transistors
64 kHz clock speed	> 3 GHz clock speed.

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Source: http://en.wikipedia.org/wiki/Image:Silicon_chip_3d.png.

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- Here: mostly MOR for linear systems, they occur in micro electronics through modified nodal analysis (MNA) for RLC networks. e.g., when
 - decoupling large linear subcircuits,
 - modeling transmission lines,
 - modeling pin packages in VLSI chips,
 - modeling circuit elements described by Maxwell's equation using partial element equivalent circuits (PEEC).

Micro Electronics/Circuit Simulation



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 \rightsquigarrow Clear need for model reduction techniques in order to facilitate or even enable circuit simulation for current and future VLSI design.

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Standard MOR techniques in circuit simulation: Krylov subspace / Padé approximation / rational interpolation methods.



Many other disciplines in computational sciences and engineering like

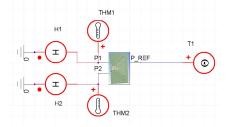
- computational fluid dynamics (CFD),
- computational electromagnetics,
- chemical process engineering,
- design of MEMS/NEMS (micro/nano-electrical-mechanical systems),
- computational acoustics,
- . . .

Motivating Examples Electro-Thermic Simulation of Integrated Circuit (IC)



[Source: Evgenii Rudnyi, CADFEM GmbH]

• $SIMPLORER^{(R)}$ test circuit with 2 transistors.



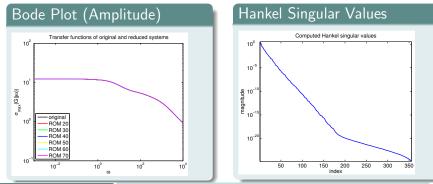
- Conservative thermic sub-system in SIMPLORER: voltage ~→ temperature, current ~→ heat flow.
- Original model: n = 270.593, m = q = 2 ⇒
 Computing time (on Intel Xeon dualcore 3GHz, 1 Thread):
 - Main computational cost for set-up data ≈ 22 min.
 - Computation of reduced models from set-up data: 44–49sec. (r = 20-70).
 - Bode plot (MATLAB on Intel Core i7, 2,67GHz, 12GB):
 7.5h for original system, < 1min for reduced system.
 - Speed-up factor: 18 including $/ \ge 450$ excluding reduced model generation!

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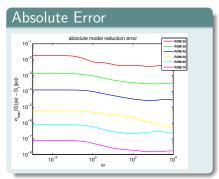
Motivating Examples

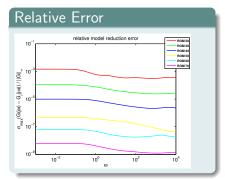
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A Nonlinear Model from Computational Neurosciences: the FitzHugh-Nagumo System

• Simple model for neuron (de-)activation [Chaturantabut/Sorensen 2009]

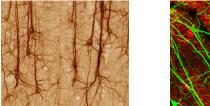
$$\epsilon v_t(x,t) = \epsilon^2 v_{xx}(x,t) + f(v(x,t)) - w(x,t) + g,$$

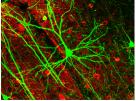
$$w_t(x,t) = hv(x,t) - \gamma w(x,t) + g,$$

with f(v) = v(v - 0.1)(1 - v) and initial and boundary conditions

$$egin{aligned} & v(x,0) = 0, & w(x,0) = 0, & x \in [0,1] \\ & v_x(0,t) = -i_0(t), & v_x(1,t) = 0, & t \geq 0, \end{aligned}$$

where $\epsilon = 0.015$, h = 0.5, $\gamma = 2$, g = 0.05, $i_0(t) = 50000t^3 \exp(-15t)$.





Source: http://en.wikipedia.org/wiki/Neuron



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where $\epsilon = 0.015$, h = 0.5, $\gamma = 2$, g = 0.05, $i_0(t) = 50000t^3 \exp(-15t)$.

- Parameter g handled as an additional input.
- Original state dimension $n = 2 \cdot 400$, QBDAE dimension $N = 3 \cdot 400$, reduced QBDAE dimension r = 26, chosen expansion point $\sigma = 1$.



A Nonlinear Model from Computational Neurosciences: the FitzHugh-Nagumo System



Nonlinear Model Reduction

Given a large-scale control-affine nonlinear control system of the form

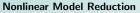
$$\Sigma : \begin{cases} \dot{x}(t) = f(x(t)) + bu(t), \\ y(t) = c^{T} x(t), \quad x(0) = x_{0}, \end{cases}$$

with $f : \mathbb{R}^n \to \mathbb{R}^n$ nonlinear and $b, c \in \mathbb{R}^n, x \in \mathbb{R}^n, u, y \in \mathbb{R}$.



$$\hat{\Sigma}: \begin{cases} \dot{\hat{x}}(t) = \hat{f}(\hat{x}(t)) + \hat{b}u(t), \\ \hat{y}(t) = \hat{c}^{\mathsf{T}}\hat{x}(t), \quad \hat{x}(0) = \hat{x}_0, \end{cases}$$

with $\hat{f}: \mathbb{R}^{\hat{n}} \to \mathbb{R}^{\hat{n}}$ and $\hat{b}, \hat{c} \in \mathbb{R}^{\hat{n}}, \, x \in \mathbb{R}^{\hat{n}}, \, u \in \mathbb{R}$ and





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Common Reduction Techniques

- Proper Orthogonal Decomposition (POD)
 - Take computed or experimental 'snapshots' of full model: $[x(t_1), x(t_2), \dots, x(t_N)] =: X$,
 - perform SVD of snapshot matrix: $X = VSW^T \approx V_{\hat{n}}S_{\hat{n}}W_{\hat{n}}^T$.
 - Reduction by POD-Galerkin projection: $\dot{\hat{x}} = V_{\hat{n}}^T f(V_{\hat{n}} \hat{x}) + V_{\hat{n}}^T Bu$.
 - Requires evaluation of f
 - → discrete empirical interpolation [Sorensen/Chaturantabut '09].
 - Input dependency due to 'snapshots'!

Introduction

Common Reduction Techniques



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Trajectory Piecewise Linear (TPWL)

- Linearize f along trajectory,
- reduce resulting linear systems,
- construct reduced model by weighted sum of linear systems.
- Requires simulation of original model and several linear reduction steps, many heuristics.





Let us start with linear systems, i.e. f(x) = Ax.

Two common system norms for measuring approximation quality:

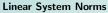
•
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-norm, $||\Sigma||_{\mathcal{H}_2} = \left(\frac{1}{2\pi}\int_0^{2\pi} \operatorname{tr}\left(H^*(-i\omega)H(i\omega)\right)d\omega\right)^{\frac{1}{2}}$,

•
$$\mathcal{H}_{\infty}$$
-norm, $||\Sigma||_{\mathcal{H}_{\infty}} = \sup_{\omega \in \mathbb{R}} \sigma_{max} \left(\mathcal{H}(i\omega) \right),$

where

$$H(s) = C \left(sI - A \right)^{-1} B$$

denotes the corresponding transfer function of the linear system.





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We focus on the first one \rightsquigarrow interpolation-based model reduction approaches.

Error system and \mathcal{H}_2 -Optimality

[Meier/Luenberger '67]



In order to find an \mathcal{H}_2 -optimal reduced system, consider the error system $H(s) - \hat{H}(s)$ which can be realized by

$$A^{err} = \begin{bmatrix} A & 0 \\ 0 & \hat{A} \end{bmatrix}, \quad B^{err} = \begin{bmatrix} B \\ \hat{B} \end{bmatrix}, \quad C^{err} = \begin{bmatrix} C & -\hat{C} \end{bmatrix}$$

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 \rightsquigarrow first-order necessary \mathcal{H}_2 -optimality conditions (SISO)

$$H(-\lambda_i) = \hat{H}(-\lambda_i),$$

$$H'(-\lambda_i) = \hat{H}'(-\lambda_i),$$

where λ_i are the poles of the reduced system $\hat{\Sigma}$.



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$$\begin{array}{ll} H(-\lambda_i)\tilde{B}_i = \hat{H}(-\lambda_i)\tilde{B}_i, & \text{for } i = 1, \dots, \hat{n}, \\ \tilde{C}_i^T H(-\lambda_i) = \tilde{C}_i^T \hat{H}(-\lambda_i), & \text{for } i = 1, \dots, \hat{n}, \\ \tilde{c}_i^T H'(-\lambda_i)\tilde{B}_i = \tilde{C}_i^T \hat{H}'(-\lambda_i)\tilde{B}_i & \text{for } i = 1, \dots, \hat{n}, \end{array}$$

where $\hat{A} = R\Lambda R^{-T}$ is the spectral decomposition of the reduced system and $\tilde{B} = \hat{B}^T R^{-T}$, $\tilde{C} = \hat{C}R$.



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 \rightsquigarrow first-order necessary \mathcal{H}_2 -optimality conditions (MIMO)

$$\begin{split} & \mathcal{H}(-\lambda_{i})\tilde{B}_{i} = \hat{\mathcal{H}}(-\lambda_{i})\tilde{B}_{i}, & \text{for } i = 1, \dots, \hat{n}, \\ & \tilde{C}_{i}^{T}\mathcal{H}(-\lambda_{i}) = \tilde{C}_{i}^{T}\hat{\mathcal{H}}(-\lambda_{i}), & \text{for } i = 1, \dots, \hat{n}, \\ & \tilde{C}_{i}^{T}\mathcal{H}'(-\lambda_{i})\tilde{B}_{i} = \tilde{C}_{i}^{T}\hat{\mathcal{H}}'(-\lambda_{i})\tilde{B}_{i} & \text{for } i = 1, \dots, \hat{n}, \\ & \text{vec}\left(I_{p}\right)^{T}\left(e_{j}e_{i}^{T}\otimes C\right)\left(-\Lambda\otimes I_{n} - I_{\hat{n}}\otimes A\right)^{-1}\left(\tilde{B}^{T}\otimes B\right)\text{vec}\left(I_{m}\right) \\ & = \text{vec}\left(I_{p}\right)^{T}\left(e_{j}e_{i}^{T}\otimes \hat{C}\right)\left(-\Lambda\otimes I_{\hat{n}} - I_{\hat{n}}\otimes \hat{A}\right)^{-1}\left(\tilde{B}^{T}\otimes \hat{B}\right)\text{vec}\left(I_{m}\right), \\ & = 1, \dots, \hat{n} \text{ and } j = 1, \dots, p. \end{split}$$

for *i*

Interpolation of the Transfer Function [GRIMME '97]



Construct reduced transfer function by Petrov-Galerkin projection $\mathcal{P} = \textit{VW}^{\textit{T}},$ i.e.

$$\hat{H}(s) = CV \left(sI - W^{T}AV \right)^{-1} W^{T}B,$$

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$$H(\sigma_i) = \hat{H}(\sigma_i)$$
 and $H'(\sigma_i) = \hat{H}'(\sigma_i)$,

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for i = 1, ..., r. \rightsquigarrow iterative algorithms (IRKA/MIRIAm) that yield \mathcal{H}_2 -optimal models.

> [Gugercin et al. '08], [Bunse-Gerstner et al. '07], [Van Dooren et al. '08]

\mathcal{H}_2 -Model Reduction for Bilinear Systems Bilinear Control Systems



Now consider $\dot{x} = Ax + g(x, u)$ with

$$g(x, u) = Bu + [N_1, \ldots, N_m] (I_m \otimes x) u,$$

i.e. bilinear control systems:

$$\Sigma: \begin{cases} \dot{x}(t) = Ax(t) + \sum_{i=1}^{m} N_i x(t) u_i(t) + Bu(t), \\ y(t) = Cx(t), \quad x(0) = x_0, \end{cases}$$

where $A, N_i \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m}, \ C \in \mathbb{R}^{p \times n}$.

$\mathcal{H}_2\text{-}\textbf{Model Reduction for Bilinear Systems}$ Bilinear Control Systems



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- Approximation of weakly nonlinear systems → Carleman linearization.
- A lot of linear concepts can be extended, e.g. transfer functions, Gramians, Lyapunov equations, ...
- An equivalent structure arises for some stochastic control systems.

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\mathcal{H}_2 -Model Reduction for Bilinear Systems Some Basic Facts



Output Characterization (SISO): Volterra series

$$y(t) = \sum_{k=1}^{\infty} \int_0^t \int_0^{t_1} \ldots \int_0^{t_{k-1}} \mathcal{K}(t_1,\ldots,t_k) u(t-t_1-\ldots-t_k) \cdots u(t-t_k) dt_k \cdots dt_1,$$

with kernels $K(t_1, \ldots, t_k) = Ce^{At_k}N_1 \cdots e^{At_2}N_1e^{At_1}B$.

$\mathcal{H}_2\text{-}\textbf{Model}$ Reduction for Bilinear Systems $_{\text{Some Basic Facts}}$



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Multivariate Laplace-transform (SISO):

$$H_k(s_1,\ldots,s_k) = C(s_k I - A)^{-1} N_1 \cdots (s_2 I - A)^{-1} N_1 (s_1 I - A)^{-1} B_1$$

\mathcal{H}_2 -Model Reduction for Bilinear Systems Some Basic Facts



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Bilinear \mathcal{H}_2 -norm (MIMO):

$$||\Sigma||_{\mathcal{H}_{2}} := \left(\operatorname{tr} \left(\sum_{k=1}^{\infty} \int_{-\infty}^{\infty} \dots \int_{-\infty}^{\infty} \frac{1}{(2\pi)^{k}} \,\overline{H_{k}(i\omega_{1},\dots,i\omega_{k})} H_{k}^{\mathsf{T}}(i\omega_{1},\dots,i\omega_{k}) \right) \right)^{\frac{1}{2}}.$$

$$[ZHANG/LAM. '02]$$

\mathcal{H}_2 -Model Reduction for Bilinear Systems \mathcal{H}_2 -Norm Computation



Lemma

Let Σ denote a bilinear system. Then, the $\mathcal{H}_2\text{-norm}$ is given as:

$$||\Sigma||_{\mathcal{H}_2}^2 = (\operatorname{vec}(I_p))^T (C \otimes C) \left(-A \otimes I - I \otimes A - \sum_{i=1}^m N_i \otimes N_i \right)^{-1} (B \otimes B) \operatorname{vec}(I_m).$$

Error System

In order to find an \mathcal{H}_2 -optimal reduced system, define the error system $\Sigma^{err} := \Sigma - \hat{\Sigma}$ as follows:

$$A^{err} = \begin{bmatrix} A & 0 \\ 0 & \hat{A} \end{bmatrix}, \quad N_i^{err} = \begin{bmatrix} N_i & 0 \\ 0 & \hat{N}_i \end{bmatrix}, \quad B^{err} = \begin{bmatrix} B \\ \hat{B} \end{bmatrix}, \quad C^{err} = \begin{bmatrix} C & -\hat{C} \end{bmatrix}.$$

[B./BREITEN '11]

Introduction \mathcal{H}_2 -Model Reduction for Bilinear Systems Nonlinear Model Reduction Numerical Examples Conclusions

\mathcal{H}_2 -Model Reduction



 \mathcal{H}_2 -Optimality Conditions

Let us assume $\hat{\Sigma}$ is given by its eigenvalue decomposition:

$$\hat{A} = R\Lambda R^{-1}, \quad \tilde{N}_i = R^{-1}\hat{N}_i R, \quad \tilde{B} = R^{-1}\hat{B}, \quad \tilde{C} = \hat{C}R.$$

\mathcal{H}_2 -Model Reduction \mathcal{H}_2 -Optimality Conditions



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\mathcal{H}_2 -Model Reduction H₂-Optimality Conditions



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= $(\operatorname{vec}(I_q))^T \left(e_j e_\ell^T \otimes \hat{C} \right) \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes \hat{A} - \sum_{i=1}^m \tilde{N}_i \otimes \hat{N}_i \right)^{-1} \left(\tilde{B} \otimes \hat{B} \right) \operatorname{vec}(I_m).$

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Where is the connection to the interpolation of transfer functions?

\mathcal{H}_2 -Model Reduction \mathcal{H}_2 -Optimality Conditions



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\mathcal{H}_2 -Model Reduction \mathcal{H}_2 -Optimality Conditions



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 $(\operatorname{vec}(I_q))^T \left(e_j e_{\ell}^T \otimes C \right) \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes A \right)^{-1} \operatorname{vec}(B\tilde{B}^T)$
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\mathcal{H}_2 -Model Reduction \mathcal{H}_2 -Optimality Conditions



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\mathcal{H}_2 -Model Reduction \mathcal{H}_2 -Optimality Conditions



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$$(\operatorname{vec}(I_q))^T \left(e_j e_\ell^T \otimes C \right) \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes A - \sum_{i=1}^m \tilde{N}_i \otimes N_i \right)^{-1} \left(\tilde{B} \otimes B \right) \operatorname{vec}(I_m)$$

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$$H(-\lambda_{\ell})\tilde{B}_{\ell}^{T} = \hat{H}(-\lambda_{\ell})\tilde{B}_{\ell}^{T}$$

 \rightsquigarrow tangential interpolation at mirror images of reduced system poles

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Note: [FLAGG 2011] shows equivalence to interpolating the Volterra series!

 \mathcal{H}_2 -Model Reduction for Bilinear Systems Nonlinear Model Reduction Numerical Examples Conclusions and Outlook References

A First Iterative Approach

Algorithm 1 Bilinear IRKA

nput:
$$A, N_i, B, C, \hat{A}, \hat{N}_i, \hat{B}, \hat{C}$$

Dutput: $A^{opt}, N_i^{opt}, B^{opt}, C^{opt}$
1: while (change in $\Lambda > \epsilon$) do
2: $R\Lambda R^{-1} = \hat{A}, \tilde{B} = R^{-1}\hat{B}, \tilde{C} = \hat{C}R, \tilde{N}_i = R^{-1}\hat{N}_iR$
3: $\operatorname{vec}(V) = \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes A - \sum_{i=1}^m \tilde{N}_i \otimes N_i\right)^{-1} \left(\tilde{B} \otimes B\right) \operatorname{vec}(I_m)$
4: $\operatorname{vec}(W) = \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes A^T - \sum_{i=1}^m \tilde{N}_i^T \otimes N_i^T\right)^{-1} \left(\tilde{C}^T \otimes C^T\right) \operatorname{vec}(I_q)$
5: $V = \operatorname{orth}(V), W = \operatorname{orth}(W)$
6: $\hat{A} = (W^T V)^{-1} W^T A V, \hat{N}_i = (W^T V)^{-1} W^T N_i V, \hat{B} = (W^T V)^{-1} W^T B, \hat{C} = C V$
7: end while
8: $A^{opt} = \hat{A} N_i^{opt} = \hat{N}_i B^{opt} = \hat{B} C^{opt} = \hat{C}$



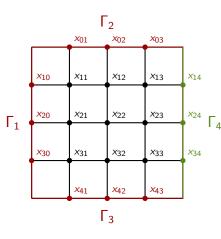
\mathcal{H}_2 -Model Reduction for Bilinear Systems



- 2-dimensional heat distribution [B./SAAK '05]
- Boundary control by spraying intensities of a cooling fluid

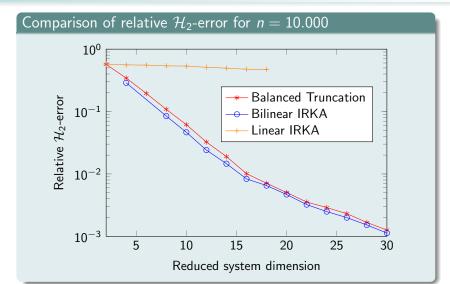
$$\begin{split} \Omega &= (0,1) \times (0,1), \\ x_t &= \Delta x & \text{ in } \Omega, \\ n \cdot \nabla x &= c \cdot u_{1,2,3}(x-1) & \text{ on } \Gamma_1, \Gamma_2, \Gamma_3, \\ x &= u_4 & \text{ on } \Gamma_4. \end{split}$$

• Spatial discretization $k \times k$ -grid $\Rightarrow \dot{x} \approx A_1 x + \sum_{i=1}^{3} N_i x u_i + B u$ $\Rightarrow A_2 = 0.$ • Output: $y = \frac{1}{k^2} \begin{bmatrix} 1 & \dots & 1 \end{bmatrix}.$



\mathcal{H}_2 -Model Reduction for Bilinear Systems A Heat Transfer Model





\mathcal{H}_2 -Model Reduction for Bilinear Systems Fokker-Planck Equation



As a second example, we consider a dragged Brownian particle whose one-dimensional motion is given by

$$dX_t = -\nabla V(X_t, t)dt + \sqrt{2\sigma}dW_t,$$

with $\sigma = \frac{2}{3}$ and $V(x, u) = W(x, t) + \Phi(x, u_t) = (x^2 - 1)^2 - xu - x$. Alternatively, one can consider ([HARTMANN ET AL. '10]),

$$\rho(x,t)dx = \mathbf{P}\left[X_t \in [x, x + dx)\right]$$

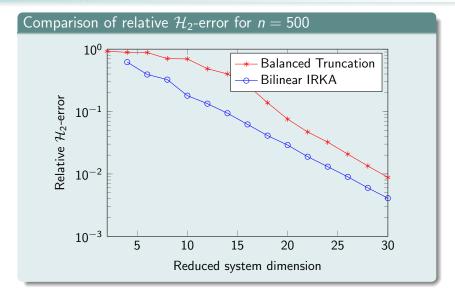
which is described by the Fokker-Planck equation

$$\begin{split} &\frac{\partial\rho}{\partial t} = \sigma\Delta\rho + \nabla\cdot(\rho\nabla V), \qquad & (x,t)\in(-2,2)\times(0,T], \\ &0 = \sigma\nabla\rho + \rho\nabla B, \qquad & (x,t)\in\{-2,2\}\times[0,T], \\ &\rho_0 = \rho, \qquad & (x,t)\in(-2,2)\times 0. \end{split}$$

Output C discrete characteristic function of the interval [0.95, 1.05].

\mathcal{H}_2 -Model Reduction for Bilinear Systems





Nonlinear Model Reduction Quadratic-Bilinear Differential Algebraic Equations (QBDAEs)



Coming back to the more general case with nonlinear f(x), we consider the class of quadratic-bilinear differential algebraic equations

$$\Sigma: \begin{cases} E\dot{x}(t) = A_1 x(t) + A_2 x(t) \otimes x(t) + N x(t) u(t) + B u(t), \\ y(t) = C x(t), \quad x(0) = x_0, \end{cases}$$

where $E, A_1, N \in \mathbb{R}^{n \times n}, A_2 \in \mathbb{R}^{n \times n^2}$ (Hessian tensor), $B, C^T \in \mathbb{R}^n$ are quite helpful.

- A large class of smooth nonlinear control-affine systems can be transformed into the above type of control system.
- The transformation is exact, but a slight increase of the state dimension has to be accepted.
- Input-output behavior can be characterized by generalized transfer functions →→ enables us to use Krylov-based reduction techniques.

Transformation via McCormick Relaxation

Theorem [Gu'09]

Assume that the state equation of a nonlinear system $\boldsymbol{\Sigma}$ is given by

$$\dot{x} = a_0 x + a_1 g_1(x) + \ldots + a_k g_k(x) + Bu,$$

where $g_i(x) : \mathbb{R}^n \to \mathbb{R}^n$ are compositions of uni-variable rational, exponential, logarithmic, trigonometric or root functions, respectively. Then, by iteratively taking derivatives and adding algebraic equations, respectively, Σ can be transformed into a system of QBDAEs.



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$$\dot{x}_1 = \exp(-x_2) \cdot \sqrt{x_1^2 + 1}, \quad \dot{x}_2 = -x_2 + u.$$



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Example

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$$z_1 := \exp(-x_2), \quad z_2 := \sqrt{x_1^2 + 1}.$$

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•
$$\dot{x}_1 = z_1 \cdot z_2$$
, $\dot{x}_2 = -x_2 + u$, $\dot{z}_1 = -z_1 \cdot (-x_2 + u)$,
 $\dot{z}_2 = \frac{2 \cdot x_1 \cdot z_1 \cdot z_2}{2 \cdot z_2} = x_1 \cdot z_1$.



Nonlinear Model Reduction

Variational Analysis and Linear Subsystems



Analysis of nonlinear systems by variational equation approach:

Nonlinear Model Reduction

Variational Analysis and Linear Subsystems



Analysis of nonlinear systems by variational equation approach:

• consider input of the form $\alpha u(t)$,

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Nonlinear Model Reduction

Variational Analysis and Linear Subsystems



Analysis of nonlinear systems by variational equation approach:

- consider input of the form $\alpha u(t)$,
- nonlinear system is assumed to be a series of homogeneous nonlinear subsystems, i.e. response should be of the form

$$x(t) = \alpha x_1(t) + \alpha^2 x_2(t) + \alpha^3 x_3(t) + \dots$$

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• comparison of terms $lpha^i, i=1,2,\ldots$ leads to series of systems

$$\begin{split} E\dot{x}_{1} &= A_{1}x_{1} + Bu, \\ E\dot{x}_{2} &= A_{1}x_{2} + A_{2}x_{1} \otimes x_{1} + Nx_{1}u, \\ E\dot{x}_{3} &= A_{1}x_{3} + A_{2} \left(x_{1} \otimes x_{2} + x_{2} \otimes x_{1} \right) + Nx_{2}u \end{split}$$

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Nonlinear Model Reduction

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 although *i*-th subsystem is coupled nonlinearly to preceding systems, linear systems are obtained if terms x_j, j < i, are interpreted as pseudo-inputs.

Nonlinear Model Reduction

Generalized Transfer Functions



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Nonlinear Model Reduction

Generalized Transfer Functions



$$H_1(s_1) = C \underbrace{(s_1 E - A_1)^{-1} B}_{G_1(s_1)},$$

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Nonlinear Model Reduction

Generalized Transfer Functions



$$\begin{split} H_1(s_1) &= C \underbrace{(s_1 E - A_1)^{-1} B}_{G_1(s_1)}, \\ H_2(s_1, s_2) &= \frac{1}{2!} C \left((s_1 + s_2) E - A_1 \right)^{-1} \left[N \left(G_1(s_1) + G_1(s_2) \right) \right. \\ &+ A_2 \left(G_1(s_1) \otimes G_1(s_2) + G_1(s_2) \otimes G_1(s_1) \right) \right], \end{split}$$

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Nonlinear Model Reduction

Generalized Transfer Functions



$$\begin{split} H_1(s_1) &= C\underbrace{(s_1E - A_1)^{-1}B}_{G_1(s_1)}, \\ H_2(s_1, s_2) &= \frac{1}{2!}C\left((s_1 + s_2)E - A_1\right)^{-1}\left[N\left(G_1(s_1) + G_1(s_2)\right) \\ &\quad +A_2\left(G_1(s_1) \otimes G_1(s_2) + G_1(s_2) \otimes G_1(s_1)\right)\right], \\ H_3(s_1, s_2, s_3) &= \frac{1}{3!}C\left((s_1 + s_2 + s_3)E - A_1\right)^{-1} \\ &\left[N(G_2(s_1, s_2) + G_2(s_2, s_3) + G_2(s_1, s_3)) \\ &\quad +A_2\left(G_1(s_1) \otimes G_2(s_2, s_3) + G_1(s_2) \otimes G_2(s_1, s_3) \\ &\quad +G_1(s_3) \otimes G_2(s_1, s_3) + G_2(s_2, s_3) \otimes G_1(s_1) \\ &\quad +G_2(s_1, s_3) \otimes G_1(s_2) + G_2(s_1, s_2) \otimes G_1(s_3))\right]. \end{split}$$

Nonlinear Model Reduction

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Characterization via Multimoments



For simplicity, focus on the first two transfer functions. For $H_1(s_1)$, choosing σ and making use of the Neumann lemma leads to

$$H_1(s_1) = \sum_{i=0}^{\infty} C \underbrace{\left((A_1 - \sigma E)^{-1} E \right)^i (A_1 - \sigma E)^{-1} B (s_1 - \sigma)^i}_{m_{s_1,\sigma}^i}.$$

Nonlinear Model Reduction

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Similarly, specifying an expansion point (au, ξ) yields

$$H_{2}(s_{1}, s_{2}) = \frac{1}{2} \sum_{i=0}^{\infty} C\left((A_{1} - (\tau + \xi)E)^{-1}E \right)^{i} (A_{1} - (\tau + \xi)E)^{-1} (s_{1} + s_{2} - \tau - \xi)^{i} \cdot \left[A_{2} \left(\sum_{j=0}^{\infty} m_{s_{1},\tau}^{j} \otimes \sum_{k=0}^{\infty} m_{s_{2},\xi}^{k} + \sum_{k=0}^{\infty} m_{s_{2},\xi}^{k} \otimes \sum_{j=0}^{\infty} m_{s_{1},\tau}^{j} \right) + N\left(\sum_{p=0}^{\infty} m_{s_{1},\tau}^{p} + \sum_{p=0}^{\infty} m_{s_{2},\xi}^{q} \right) \right]$$

Nonlinear Model Reduction

Constructing the Projection Matrix



$$\begin{array}{l} \mbox{Goal:} \ \frac{\partial}{\partial s_1^{q-1}} H_1(\sigma) = \frac{\partial}{\partial s_1^{q-1}} \hat{H}_1(\sigma), \ \ \frac{\partial}{\partial s_1' s_2^m} H_2(\sigma,\sigma) = \frac{\partial}{\partial s_1' s_2^m} \hat{H}_2(\sigma,\sigma), \ l+m \leq q-1. \\ \mbox{Construct the following sequence of nested Krylov subspaces} \end{array}$$

Nonlinear Model Reduction

Constructing the Projection Matrix



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$$V_1 = \mathcal{K}_q \left((A_1 - \sigma E)^{-1} E, (A_1 - \sigma E)^{-1} b \right)$$

Nonlinear Model Reduction

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$$V_{1} = \mathcal{K}_{q} \left((A_{1} - \sigma E)^{-1} E, (A_{1} - \sigma E)^{-1} b \right)$$

for $i = 1 : q$
$$V_{2}^{i} = \mathcal{K}_{q-i+1} \left((A_{1} - 2\sigma E)^{-1} E, (A_{1} - 2\sigma E)^{-1} N V_{1}(:, i) \right),$$

Nonlinear Model Reduction

Constructing the Projection Matrix



Goal:
$$\frac{\partial}{\partial s_1^{q-1}} H_1(\sigma) = \frac{\partial}{\partial s_1^{q-1}} \hat{H}_1(\sigma), \quad \frac{\partial}{\partial s_1^{l} s_2^m} H_2(\sigma, \sigma) = \frac{\partial}{\partial s_1^{l} s_2^m} \hat{H}_2(\sigma, \sigma), \quad l+m \le q-1.$$

Construct the following sequence of nested Krylov subspaces
 $V_1 = \mathcal{K}_q \left((A_1 - \sigma E)^{-1} E, (A_1 - \sigma E)^{-1} b \right)$
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for $j = 1 : \min(q - i + 1, i)$
 $V_3^{i,j} = \mathcal{K}_{q-i-j+2} \left((A_1 - 2\sigma E)^{-1} E, (A_1 - 2\sigma E)^{-1} A_2 V_1(:, i) \otimes V_1(:, .) \right)$

 $V_1(:, i)$ denoting the i-th column of V_1 .

Nonlinear Model Reduction

Constructing the Projection Matrix



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$$\frac{\partial}{\partial s_1^{q-1}} H_1(\sigma) = \frac{\partial}{\partial s_1^{q-1}} \hat{H}_1(\sigma), \quad \frac{\partial}{\partial s_1^{l} s_2^{m}} H_2(\sigma, \sigma) = \frac{\partial}{\partial s_1^{l} s_2^{m}} \hat{H}_2(\sigma, \sigma), \quad l+m \le q-1.$$

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for $j = 1 : \min(q - i + 1, i)$
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 $V_1(:, i)$ denoting the i-*th* column of V_1 . Set $\mathcal{V} = \text{orth} [V_1, V_2^i, V_3^{i,j}]$ and construct $\hat{\Sigma}$ by the Galerkin-Projection $\mathcal{P} = \mathcal{V}\mathcal{V}^T$:

$$\hat{A}_1 = \mathcal{V}^T A_1 \mathcal{V} \in \mathbb{R}^{\hat{n} imes \hat{n}}, \quad \hat{A}_2 = \mathcal{V}^T A_2 (\mathcal{V} \otimes \mathcal{V}) \in \mathbb{R}^{\hat{n} imes \hat{n}^2},$$

 $\hat{N} = \mathcal{V}^T N \mathcal{V} \in \mathbb{R}^{\hat{n} imes \hat{n}}, \quad \hat{b} = \mathcal{V}^T b \in \mathbb{R}^{\hat{n}}, \quad \hat{c}^T = c^T \mathcal{V} \in \mathbb{R}^{\hat{n}}.$

Nonlinear Model Reduction



Tensors and Matricizations: A Short Excursion

[Kolda/Bader '09, Grasedyck '10]

A tensor is a vector

$$(A_i)_{i\in\mathcal{I}}\in\mathbb{R}^{\mathcal{I}}$$

indexed by a product index set

$$\mathcal{I} = \mathcal{I}_1 \times \cdots \times \mathcal{I}_d, \quad \# \mathcal{I}_j = n_j.$$

Nonlinear Model Reduction



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For a given tensor A, the *t*-matricization $A^{(t)}$ is defined as

$$A^{(t)} \in \mathbb{R}^{\mathcal{I}_t imes \mathcal{I}_{t'}}, \quad A^{(t)}_{(i_{\mu})\mu \in t, \ (i_{\mu})\mu \in t'} := A_{(i_1, \dots, i_d)}, \quad t' := \{1, \dots, d\} \setminus t.$$

Nonlinear Model Reduction



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Example: For a given 3-tensor $A_{(i_1,i_2,i_3)}$ with $i_1, i_2, i_3 \in \{1,2\}$, we have:

$$A^{(1)} = \begin{bmatrix} A_{(1,1,1)} & A_{(1,2,1)} & A_{(1,1,2)} & A_{(1,2,2)} \\ A_{(2,1,1)} & A_{(2,2,1)} & A_{(2,1,2)} & A_{(2,2,2)} \end{bmatrix},$$
$$A^{(2)} = \begin{bmatrix} A_{(1,1,1)} & A_{(2,1,1)} & A_{(1,1,2)} & A_{(2,1,2)} \\ A_{(1,2,1)} & A_{(2,2,1)} & A_{(1,2,2)} & A_{(2,2,2)} \end{bmatrix}.$$

Nonlinear Model Reduction

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Tensors and Matricizations: A Short Excursion

[Kolda/Bader '09, Grasedyck '10]

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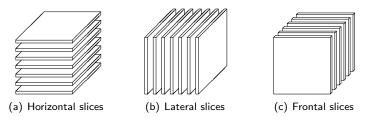


Figure : Slices of a 3rd-order tensor. [Courtesy of Tammy Kolda]

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Tensors and Matricizations: A Short Excursion [Kolda/Bader '09, Grasedyck '10]

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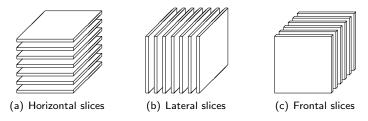


Figure : Slices of a 3rd-order tensor. [Courtesy of Tammy Kolda]

 \rightsquigarrow Allows to compute matrix products more efficiently.

Nonlinear Model Reduction

Two-Sided Projection Methods



Similarly to the linear case, one can exploit duality concepts, in order to construct two-sided projection methods.

Nonlinear Model Reduction

Two-Sided Projection Methods



Similarly to the linear case, one can exploit duality concepts, in order to construct two-sided projection methods.

Interpreting $\mathcal{A}^{(2)}$ now as the 2-matricization of the Hessian 3-tensor corresponding to A_2 , one can show that the dual Krylov spaces have to be constructed as follows

$$W_{1} = \mathcal{K}_{q} \left((A_{1} - 2\sigma E)^{-T} E^{T}, (A_{1} - 2\sigma E)^{-T} c \right)$$

for $i = 1 : q$
$$W_{2}^{i} = \mathcal{K}_{q-i+1} \left((A_{1} - \sigma E)^{-T} E^{T}, (A_{1} - \sigma E)^{-T} N^{T} W_{1}(:, i) \right),$$

for $j = 1 : \min(q - i + 1, i)$
$$W_{3}^{i,j} = \mathcal{K}_{q-i-j+2} \left((A_{1} - \sigma E)^{-T} E^{T}, (A_{1} - \sigma E)^{-T} \mathcal{A}^{(2)} V_{1}(:, i) \otimes W_{1}(:, j) \right),$$

Nonlinear Model Reduction

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$$W_{1} = \mathcal{K}_{q} \left((A_{1} - 2\sigma E)^{-T} E^{T}, (A_{1} - 2\sigma E)^{-T} c \right)$$

for $i = 1 : q$
$$W_{2}^{i} = \mathcal{K}_{q-i+1} \left((A_{1} - \sigma E)^{-T} E^{T}, (A_{1} - \sigma E)^{-T} N^{T} W_{1}(:, i) \right),$$

for $j = 1 : \min(q - i + 1, i)$
$$W_{3}^{i,j} = \mathcal{K}_{q-i-j+2} \left((A_{1} - \sigma E)^{-T} E^{T}, (A_{1} - \sigma E)^{-T} \mathcal{A}^{(2)} V_{1}(:, i) \otimes W_{1}(:, j) \right),$$

Note: Due to the symmetry of the Hessian tensor, the 3-matricization $\mathcal{A}^{(3)}$ coincides with $\mathcal{A}^{(2)}$.

Nonlinear Model Reduction

Multimoment matching



Theorem

• $\Sigma = (E, A_1, A_2, N, b, c)$ original QBDAE system.

• Reduced system by Petrov-Galerkin projection $\mathcal{P} = \mathcal{V} \mathcal{W}^{\mathcal{T}}$ with

$$W_1 = \mathcal{K}_{q_1}(E, A_1, b, \sigma), \quad W_1 = \mathcal{K}_{q_1}(E^T, A_1^T, c, 2\sigma)$$

$$i = 1 : q_2$$

$$V_2 = \mathcal{K}_{q_2 - i + 1} (E, A_1, NV_1(:, i), 2\sigma)$$

$$W_2 = \mathcal{K}_{q_2 - i + 1} (E^T, A_1^T, N^T W_1(:, i), \sigma)$$
for $j = 1 : \min(q_2 - i + 1, i)$

$$V_3 = \mathcal{K}_{q_2 - i - j + 2} (E, A_1, A_2 V_1(:, i) \otimes V_1(:, j), 2\sigma)$$

$$W_3 = \mathcal{K}_{q_2-i-j+2}\left(E^T, \mathcal{A}_1^T, \mathcal{A}^{(2)}V_1(:,i)\otimes W_1(:,j), \sigma\right).$$

Then, it holds:

for

$$\frac{\partial^{i} H_{1}}{\partial s_{1}^{i}}(\sigma) = \frac{\partial^{i} \hat{H}_{1}}{\partial s_{1}^{i}}(\sigma), \quad \frac{\partial^{i} H_{1}}{\partial s_{1}^{i}}(2\sigma) = \frac{\partial^{i} \hat{H}_{1}}{\partial s_{1}^{i}}(2\sigma), \quad i = 0, \dots, q_{1} - 1,$$

$$\frac{\partial^{i+j}}{\partial s_{1}^{i} s_{2}^{j}} H_{2}(\sigma, \sigma) = \frac{\partial^{i+j}}{\partial s_{1}^{i} s_{2}^{j}} \hat{H}_{2}(\sigma, \sigma), \qquad i+j \leq 2q_{2} - 1.$$

 \mathcal{H}_2 -Model Reduction for Bilinear Systems Nonlinear Model Reduction Numerical Examples Conclusions and Outlook References

Numerical Examples

Two-Dimensional Burgers Equation



• 2D-Burgers equation on
$$(0,1) \times (0,1) \times [0,T]$$

 $u_t = -(u \cdot \nabla) u + \nu \Delta u$

with $u(x, y, t) \in \mathbb{R}^2$ describing the motion of a compressible fluid.

 $:=\Omega$

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Consider initial and boundary conditions

$$\begin{array}{ll} u_x(x,y,0) = \frac{\sqrt{2}}{2}, & u_y(x,y,0) = \frac{\sqrt{2}}{2}, & \text{for } (x,y) \in \Omega_1 := (0,0.5], \\ u_x(x,y,0) = 0, & u_y(x,y,0) = 0, & \text{for } (x,y) \in \Omega \backslash \Omega_1, \\ u_x = 0, & u_y = 0, & \text{for } (x,y) \in \partial \Omega. \end{array}$$

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Two-Dimensional Burgers Equation



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• Spatial discretization \rightsquigarrow QBDAE system with nonzero I.C. and $N = 0 \rightsquigarrow$ reformulate as system with zero I.C. and constant input.

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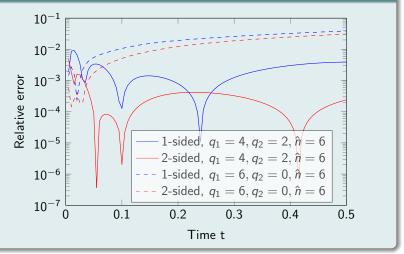
- Spatial discretization \rightsquigarrow QBDAE system with nonzero I.C. and $N = 0 \rightsquigarrow$ reformulate as system with zero I.C. and constant input.
- Output *C* chosen to be average *x*-velocity.

Numerical Examples

Two-Dimensional Burgers Equation



Comparison of relative time-domain error for n = 1600



 \mathcal{H}_2 -Model Reduction for Bilinear Systems Nonlinear Model Reduction Numerical Examples Conclusions and Outlook References

Numerical Examples

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Now consider initial and boundary conditions

$$\begin{array}{ll} u_x(x,y,0)=0, & u_y(x,y,0)=0, & \mbox{ for } x,y\in\Omega, \\ u_x=\cos(\pi t), & u_y=\cos(2\pi t), & \mbox{ for } (x,y)\in\{0,1\}\times(0,1), \\ u_x=\sin(\pi t), & u_y=\sin(2\pi t), & \mbox{ for } (x,y)\in(0,1)\times\{0,1\}. \end{array}$$

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• Spatial discretization \rightsquigarrow QBDAE system with zero I.C. and 4 inputs $B \in \mathbb{R}^{n \times 4}$, N_1, N_2, N_3, N_4 , ROM with $q_1 = 5, q_2 = 2, \sigma = 0, \hat{n} = 52$.

 \mathcal{H}_2 -Model Reduction for Bilinear Systems Nonlinear Model Reduction Numerical Examples Conclusions and Outlook References

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- State reconstruction by reduced model $x \approx V\hat{x}$, max. rel. err < 3%.

 \mathcal{H}_2 -Model Reduction for Bilinear Systems Nonlinear Model Reduction Numerical Examples Conclusions and Outlook References

Numerical Examples

The Chafee-Infante equation



• Consider PDE with a cubic nonlinearity:

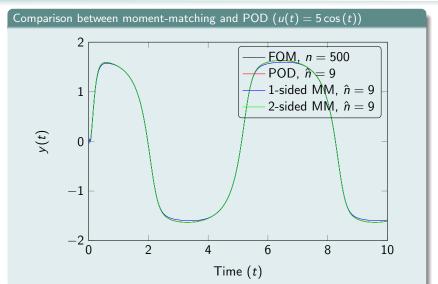
$$\begin{split} v_t + v^3 &= v_{xx} + v, & \text{ in } (0,1) \times (0,T), \\ v(0,\cdot) &= u(t), & \text{ in } (0,T), \\ v_x(1,\cdot) &= 0, & \text{ in } (0,T), \\ v(x,0) &= v_0(x), & \text{ in } (0,1) \end{split}$$

• original state dimension n = 500, QBDAE dimension $N = 2 \cdot 500$, reduced QBDAE dimension r = 9

Numerical Examples



The Chafee-Infante equation

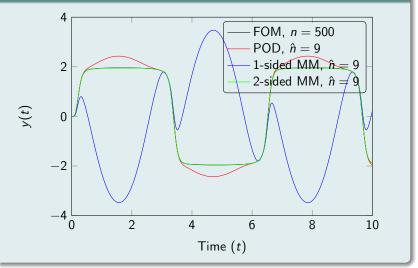


Numerical Examples



The Chafee-Infante equation





Numerical Examples

The FitzHugh-Nagumo System



• FitzHugh-Nagumo system modeling a neuron

[Chaturantabut, Sorensen '09]

$$\begin{aligned} \epsilon v_t(x,t) &= \epsilon^2 v_{xx}(x,t) + f(v(x,t)) - w(x,t) + g, \\ w_t(x,t) &= hv(x,t) - \gamma w(x,t) + g, \end{aligned}$$

with f(v) = v(v - 0.1)(1 - v) and initial and boundary conditions

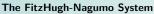
$$egin{aligned} &v(x,0)=0, &w(x,0)=0, &x\in[0,1],\ &v_x(0,t)=-i_0(t), &v_x(1,t)=0, &t\geq 0, \end{aligned}$$

where

 $\epsilon = 0.015, \ h = 0.5, \ \gamma = 2, \ g = 0.05, \ i_0(t) = 5 \cdot 10^4 t^3 \exp(-15t)$

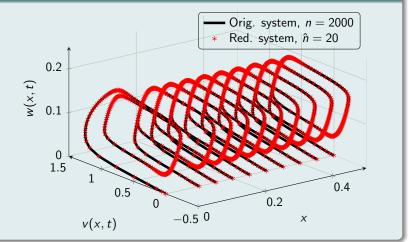
• original state dimension $n = 2 \cdot 1000$, QBDAE dimension $N = 3 \cdot 1000$, reduced QBDAE dimension r = 20

Numerical Examples







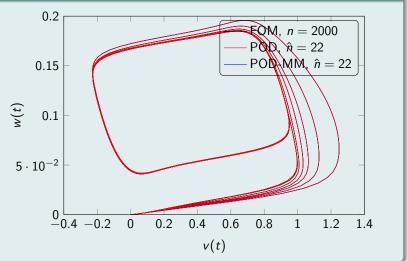


Numerical Examples

The FitzHugh-Nagumo System



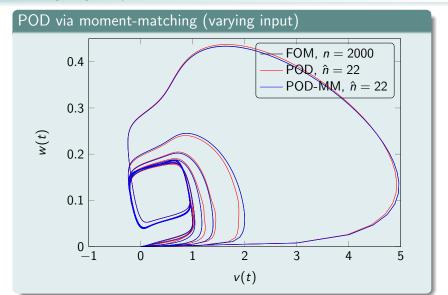




Numerical Examples

The FitzHugh-Nagumo System





Conclusions and Outlook

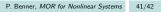
- Many nonlinear dynamics can be expressed by a system of quadratic-bilinear differential algebraic equations.
- For this type of systems, a frequency domain analysis leads to certain generalized transfer functions.
- There exist Krylov subspace methods that extend the concept of moment-matching → using basic tools from tensor theory allows for better approximations.
- In contrast to other methods like TPWL and POD, the reduction process is independent of the control input.



Conclusions and Outlock

Conclusions and Outlook

- Many nonlinear dynamics can be expressed by a system of quadratic-bilinear differential algebraic equations.
- For this type of systems, a frequency domain analysis leads to certain generalized transfer functions.
- There exist Krylov subspace methods that extend the concept of moment-matching → using basic tools from tensor theory allows for better approximations.
- In contrast to other methods like TPWL and POD, the reduction process is independent of the control input.
- Optimal choice of interpolation points?
- Stability/index-preserving reduction possible?





References





P. Benner and T. Breiten.

Interpolation-Based H_2 -Model Reduction of Bilinear Control Systems. SIAM JOURNAL ON MATRIX ANALYSIS AND APPLICATIONS, 33(3):859–885, 2012.



P. Benner and T. Breiten.

Krylov-Subspace Based Model Reduction of Nonlinear Circuit Models Using Bilinear and Quadratic-Linear Approximations.

In M. Günther, A. Bartel, M. Brunk, S. Schöps, M. Striebel (Eds.), *Progress in Industrial Mathematics at ECMI 2010*, MATHEMATICS IN INDUSTRY, 17:153–159, Springer-Verlag, Berlin, 2012.



P. Benner and T. Breiten.

Two-Sided Moment Matching Methods for Nonlinear Model Reduction. MPI MAGDEBURG PREPRINT MPIMD/12-12, June 2012.



P. Benner and T. Damm.

Lyapunov Equations, Energy Functionals, and Model Order Reduction of Bilinear and Stochastic Systems.

SIAM JOURNAL ON CONTROL AND OPTIMIZTION, 49(2):686-711, 2011.



C. Gu.

QLMOR: A Projection-Based Nonlinear Model Order Reduction Approach Using Quadratic-Linear Representation of Nonlinear Systems.

IEEE TRANSACTIONS ON COMPUTER-AIDED DESIGN OF INTEGRATED CIRCUITS AND SYSTEMS, 30(9):1307–1320, 2011.