

MODEL REDUCTION FOR NONLINEAR SYSTEMS WITHOUT SNAPSHOTS

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Overview



- Introduction
 - Model Reduction for Dynamical Systems
 - Application Areas
 - Motivating Examples
 - Nonlinear Model Reduction
- 2 \mathcal{H}_2 -Model Reduction for Bilinear Systems
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 - QBDAE Systems
 - Numerical Examples
- Conclusions and Outlook

Introduction

Model Reduction for Dynamical Systems

Dynamical Systems

$$\Sigma : \left\{ \begin{array}{lcl} \dot{x}(t) & = & f(t, x(t), u(t)), & x(t_0) = x_0, \\ y(t) & = & g(t, x(t), u(t)) \end{array} \right.$$

with

- states $x(t) \in \mathbb{R}^n$,
- inputs $u(t) \in \mathbb{R}^m$,
- outputs $y(t) \in \mathbb{R}^p$.



Model Reduction for Dynamical Systems



Original System

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Goal

 $\|y - \hat{y}\| < \text{tolerance} \cdot \|u\|$ for all admissible input signals.

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Reduced-Order Model (ROM)

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- states $\hat{x}(t) \in \mathbb{R}^r$, $r \ll n$
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Goal:

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Secondary goal: reconstruct approximation of x from \hat{x} .

Application Areas



Historically:

- structural dynamics (1960ies: modal truncation, Guyan/Craig-Bampton reduction),
- computational fluid dynamics/CFD (1970/80ies: proper orthogonal decomposition/POD),
- control design (1980ies: balanced truncation),
- microelectronics/circuit simulation (1990ies: moment matching/Padé approximation [Freund!]).

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Now: many other disciplines in computational sciences and engineering like

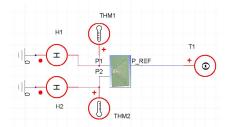
- design of MEMS/NEMS (micro/nano-electrical-mechanical systems),
- computational electromagnetics,
- computational neurosciences,
- computational (insert your favorite discipline here),
- chemical process engineering,
- biomedical engineering,
- **a**

[Source: Evgenii Rudnyi, CADFEM GmbH]

Motivating Examples

Electro-Thermic Simulation of Integrated Circuit (IC)

• SIMPLORER® test circuit with 2 transistors.



- Conservative thermic sub-system in SIMPLORER: voltage → temperature, current → heat flow.
- Original model: n = 270.593, $m = q = 2 \Rightarrow$ Computing time (on Intel Xeon dualcore 3GHz, 1 Thread):
 - Main computational cost for set-up data $\approx 22 \text{min}.$
 - Computation of reduced models from set-up data: 44–49sec. (r = 20-70).
 - Bode plot (MATLAB on Intel Core i7, 2,67GHz, 12GB):
 7.5h for original system, < 1min for reduced system.
 - Speed-up factor: 18 including / ≥ 450 excluding reduced model generation!

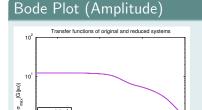
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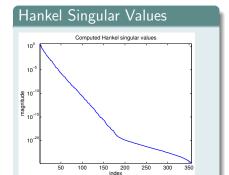
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ROM 70

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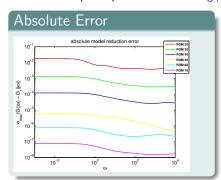
P. Benner, MOR for Nonlinear Systems

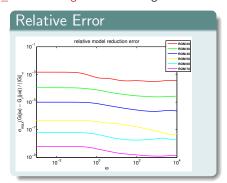
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Motivating Examples

A Nonlinear Model from Computational Neurosciences: the FitzHugh-Nagumo System

• Simple model for neuron (de-)activation [Chaturantabut/Sorensen 2009]

$$\epsilon v_t(x,t) = \epsilon^2 v_{xx}(x,t) + f(v(x,t)) - w(x,t) + g,$$

$$w_t(x,t) = hv(x,t) - \gamma w(x,t) + g,$$

with f(v) = v(v - 0.1)(1 - v) and initial and boundary conditions

$$v(x,0)=0,$$

$$w(x,0)=0, \qquad x\in [0,1]$$

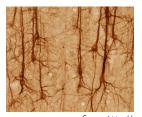
$$\mathsf{x} \in [0,1]$$

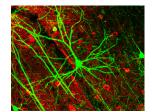
$$v_{x}(0,t)=-i_{0}(t), \qquad v_{x}(1,t)=0,$$

$$v_{\mathsf{x}}(1,t)=0$$

$$t\geq 0$$
,

where $\epsilon = 0.015, h = 0.5, \gamma = 2, g = 0.05, i_0(t) = 50000t^3 \exp(-15t)$.





Source: http://en.wikipedia.org/wiki/Neuron

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$$v(x,0) = 0,$$
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 $v_x(0,t) = -i_0(t),$ $v_x(1,t) = 0,$ $t \ge 0,$

where
$$\epsilon = 0.015$$
, $h = 0.5$, $\gamma = 2$, $g = 0.05$, $i_0(t) = 50000t^3 \exp(-15t)$.

- Parameter g handled as an additional input.
- Original state dimension $n = 2 \cdot 400$, QBDAE dimension $N = 3 \cdot 400$, reduced QBDAE dimension r = 26, chosen expansion point $\sigma = 1$.

Motivating Examples

A Nonlinear Model from Computational Neurosciences: the FitzHugh-Nagumo System

Introduction

Nonlinear Model Reduction



Given a large-scale control-affine nonlinear control system of the form

$$\Sigma: \begin{cases} \dot{x}(t) = f(x(t)) + bu(t), \\ y(t) = c^{T}x(t), \quad x(0) = x_0, \end{cases}$$

with $f: \mathbb{R}^n \to \mathbb{R}^n$ nonlinear and $b, c \in \mathbb{R}^n$, $x \in \mathbb{R}^n$, $u, y \in \mathbb{R}$.



$$\hat{\Sigma}: \left\{ \begin{aligned} \hat{\hat{x}}(t) &= \hat{f}(\hat{x}(t)) + \hat{b}u(t), \\ \hat{y}(t) &= \hat{c}^T \hat{x}(t), \quad \hat{x}(0) = \hat{x}_0, \end{aligned} \right\}$$

with $\hat{f}: \mathbb{R}^{\hat{n}} \to \mathbb{R}^{\hat{n}}$ and $\hat{b}, \hat{c} \in \mathbb{R}^{\hat{n}}, x \in \mathbb{R}^{\hat{n}}, u \in \mathbb{R}$ and



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Common Reduction Techniques

Proper Orthogonal Decomposition (POD)

- Take computed or experimental 'snapshots' of full model: $[x(t_1), x(t_2), \dots, x(t_N)] =: X$,
- perform SVD of snapshot matrix: $X = VSW^T \approx V_{\hat{n}}S_{\hat{n}}W_{\hat{n}}^T$.
- Reduction by POD-Galerkin projection: $\dot{\hat{x}} = V_{\hat{n}}^T f(V_{\hat{n}} \hat{x}) + V_{\hat{n}}^T Bu$.
- Requires evaluation of *f*
 - $\rightsquigarrow discrete\ empirical\ interpolation\ [Sorensen/Chaturantabut\ '09].$
- Input dependency due to 'snapshots'!

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Trajectory Piecewise Linear (TPWL)

- Linearize f along trajectory,
- reduce resulting linear systems,
- construct reduced model by weighted sum of linear systems.
- Requires simulation of original model and several linear reduction steps, many heuristics.

Introduction Linear System Norms



First consider linear systems, i.e. $f(x) = Ax \rightsquigarrow$

$$\dot{x}(t) = Ax(t) + Bu(t), \ y(t) = Cx(t) \simeq Y(s) = C(sI - A)^{-1}BU(s).$$

Two common system norms for measuring approximation quality:

•
$$\mathcal{H}_2$$
-norm, $\|\Sigma\|_{\mathcal{H}_2} = \left(\frac{1}{2\pi} \int_0^{2\pi} \operatorname{tr}\left(H^*(-i\omega)H(i\omega)\right) d\omega\right)^{\frac{1}{2}}$,

$$\bullet \ \mathcal{H}_{\infty}\text{-norm, } \|\Sigma\|_{\mathcal{H}_{\infty}} = \sup_{\omega \in \mathbb{R}} \sigma_{\max}\left(H(i\omega)\right),$$

where

$$H(s) = C (sI - A)^{-1} B$$

denotes the corresponding transfer function of the linear system.



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We focus on the first one \rightsquigarrow interpolation-based model reduction approaches.

Error system and \mathcal{H}_2 -Optimality

[Meier/Luenberger '67]

In order to find an \mathcal{H}_2 -optimal reduced system, consider the error system $H(s) - \hat{H}(s)$ which can be realized by

$$A^{err} = egin{bmatrix} A & 0 \ 0 & \hat{A} \end{bmatrix}, \quad B^{err} = egin{bmatrix} B \ \hat{B} \end{bmatrix}, \quad C^{err} = egin{bmatrix} C & -\hat{C} \end{bmatrix}.$$

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Assuming a coordinate system in which \hat{A} is diagonal and taking derivatives of

$$\|H(.) - \hat{H}(.)\|_{\mathcal{H}_2}^2$$

with respect to free parameters in $\Lambda(\hat{A}), \hat{B}, \hat{C} \rightsquigarrow$ first-order necessary \mathcal{H}_2 -optimality conditions (SISO)

$$H(-\hat{\lambda}_i) = \hat{H}(-\hat{\lambda}_i),$$

$$H'(-\hat{\lambda}_i) = \hat{H}'(-\hat{\lambda}_i).$$

where $\hat{\lambda}_i$ are the poles of the reduced system $\hat{\Sigma}$.

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First-order necessary \mathcal{H}_2 -optimality conditions (MIMO):

$$\begin{split} H(-\hat{\lambda}_i)\tilde{B}_i &= \hat{H}(-\hat{\lambda}_i)\tilde{B}_i, & \text{for } i = 1, \dots, \hat{n}, \\ \tilde{C}_i^T H(-\hat{\lambda}_i) &= \tilde{C}_i^T \hat{H}(-\hat{\lambda}_i), & \text{for } i = 1, \dots, \hat{n}, \\ \tilde{C}_i^T H'(-\hat{\lambda}_i)\tilde{B}_i &= \tilde{C}_i^T \hat{H}'(-\hat{\lambda}_i)\tilde{B}_i & \text{for } i = 1, \dots, \hat{n}, \end{split}$$

where $\hat{A} = R\hat{\Lambda}R^{-T}$ is the spectral decomposition of the reduced system and $\tilde{B} = \hat{B}^T R^{-T}$. $\tilde{C} = \hat{C}R$.

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Error system and \mathcal{H}_2 -Optimality

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$$\tilde{C}_{i}^{T}H(-\hat{\lambda}_{i}) = \tilde{C}_{i}^{T}\hat{H}(-\hat{\lambda}_{i}), \qquad \text{for } i = 1, \dots, \hat{n},$$

$$\tilde{C}_{i}^{T}H'(-\hat{\lambda}_{i})\tilde{B}_{i} = \tilde{C}_{i}^{T}\hat{H}'(-\hat{\lambda}_{i})\tilde{B}_{i} \qquad \text{for } i = 1, \dots, \hat{n},$$

$$\Leftrightarrow \text{vec}(I_{p})^{T}\left(e_{j}e_{i}^{T} \otimes C\right)\left(-\hat{\Lambda} \otimes I_{n} - I_{\hat{n}} \otimes A\right)^{-1}\left(\tilde{B}^{T} \otimes B\right)\text{vec}(I_{m})$$

$$= \text{vec}(I_{p})^{T}\left(e_{j}e_{i}^{T} \otimes \hat{C}\right)\left(-\hat{\Lambda} \otimes I_{\hat{n}} - I_{\hat{n}} \otimes \hat{A}\right)^{-1}\left(\tilde{B}^{T} \otimes \hat{B}\right)\text{vec}(I_{m}),$$

$$\text{for } i = 1, \dots, \hat{n} \text{ and } j = 1, \dots, p.$$

Introduction



Interpolation of the Transfer Function [GRIMME '97]

Construct reduced transfer function by Petrov-Galerkin projection $\mathcal{P} = VW^T$, i.e.

$$\hat{H}(s) = CV (sI - W^{T}AV)^{-1} W^{T}B,$$

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where V and W are given as

$$V = [(\sigma_1 I - A)^{-1} B, \dots, (\sigma_r I - A)^{-1} B],$$

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(C)

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Then

$$H(\sigma_i) = \hat{H}(\sigma_i)$$
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Introduction

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Starting with an initial guess for $\hat{\Lambda}$ and setting $\sigma_i \equiv -\hat{\lambda}_i \rightsquigarrow$ iterative algorithms (IRKA/MIRIAm) that yield \mathcal{H}_2 -optimal models.

[Gugercin et al. '06/'08], [Bunse-Gerstner et al. '07], [Van Dooren et al. '08]

\mathcal{H}_2 -Model Reduction for Bilinear Systems

Bilinear Control Systems

Now consider $\dot{x} = Ax + g(x, u)$ with

$$g(x, u) = Bu + [N_1, \dots, N_m] (I_m \otimes x) u,$$

i.e. bilinear control systems:

$$\Sigma: \begin{cases} \dot{x}(t) = Ax(t) + \sum_{i=1}^{m} N_i x(t) u_i(t) + Bu(t), \\ y(t) = Cx(t), \quad x(0) = x_0, \end{cases}$$

where $A, N_i \in \mathbb{R}^{n \times n}, B \in \mathbb{R}^{n \times m}, C \in \mathbb{R}^{p \times n}$.

H₂-Model Reduction for Bilinear Systems



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where $A, N_i \in \mathbb{R}^{n \times n}, B \in \mathbb{R}^{n \times m}, C \in \mathbb{R}^{p \times n}$.

- A lot of linear concepts can be extended, e.g. moment matching [Bai!], Gramians, Lyapunov equations, . . .
- An equivalent structure arises for some stochastic control systems.





Some Basic Facts

Output Characterization (SISO): Volterra series

$$y(t) = \sum_{k=1}^{\infty} \int_{0}^{t} \int_{0}^{t_{1}} \dots \int_{0}^{t_{k-1}} K(t_{1}, \dots, t_{k}) u(t-t_{1}-\dots-t_{k}) \cdots u(t-t_{k}) dt_{k} \cdots dt_{1},$$

with kernels $K(t_1,\ldots,t_k)=Ce^{At_k}N_1\cdots e^{At_2}N_1e^{At_1}B$.

\mathcal{H}_2 -Model Reduction for Bilinear Systems



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with kernels $K(t_1, \ldots, t_k) = Ce^{At_k} N_1 \cdots e^{At_2} N_1 e^{At_1} B$.

Multivariate Laplace-transform:

$$H_k(s_1,\ldots,s_k) = C(s_kI-A)^{-1}N_1\cdots(s_2I-A)^{-1}N_1(s_1I-A)^{-1}B.$$

Some Basic Facts

H₂-Model Reduction for Bilinear Systems



Output Characterization (SISO): Volterra series

$$y(t) = \sum_{k=1}^{\infty} \int_0^t \int_0^{t_1} \dots \int_0^{t_{k-1}} K(t_1, \dots, t_k) u(t-t_1-\dots-t_k) \cdots u(t-t_k) dt_k \cdots dt_1,$$

with kernels $K(t_1, \ldots, t_k) = Ce^{At_k} N_1 \cdots e^{At_2} N_1 e^{At_1} B$.

Multivariate Laplace-transform:

$$H_k(s_1,\ldots,s_k)=C(s_kI-A)^{-1}N_1\cdots(s_2I-A)^{-1}N_1(s_1I-A)^{-1}B.$$

Bilinear \mathcal{H}_2 -norm:

Some Basic Facts

$$||\Sigma||_{\mathcal{H}_2} := \left(\operatorname{tr} \left(\sum_{k=1}^{\infty} \int_{-\infty}^{\infty} \dots \int_{-\infty}^{\infty} \frac{1}{(2\pi)^k} \, \overline{H_k(i\omega_1, \dots, i\omega_k)} H_k^T(i\omega_1, \dots, i\omega_k) \right) \right)^{\frac{1}{2}}.$$
[ZHANG/LAM. '02]



 \mathcal{H}_2 -Norm Computation

Lemma

[B./Breiten '11]

Let Σ denote a bilinear system. Then, the \mathcal{H}_2 -norm is given as:

$$||\Sigma||_{\mathcal{H}_2}^2 = (\operatorname{vec}(I_p))^T (C \otimes C) \left(-A \otimes I - I \otimes A - \sum_{i=1}^m N_i \otimes N_i \right)^{-1} (B \otimes B) \operatorname{vec}(I_m).$$

Error System

In order to find an \mathcal{H}_2 -optimal reduced system, define the error system $\Sigma^{err} := \Sigma - \hat{\Sigma}$ as follows:

$$A^{err} = \begin{bmatrix} A & 0 \\ 0 & \hat{A} \end{bmatrix}, \quad N_i^{err} = \begin{bmatrix} N_i & 0 \\ 0 & \hat{N}_i \end{bmatrix}, \quad B^{err} = \begin{bmatrix} B \\ \hat{B} \end{bmatrix}, \quad C^{err} = \begin{bmatrix} C & -\hat{C} \end{bmatrix}.$$

\mathcal{H}_2 -Model Reduction

\mathcal{H}_2 -Optimality Conditions

Let us assume $\hat{\Sigma}$ is given by its eigenvalue decomposition:

$$\hat{A} = R \Lambda R^{-1}, \quad \tilde{N}_i = R^{-1} \hat{N}_i R, \quad \tilde{B} = R^{-1} \hat{B}, \quad \tilde{C} = \hat{C} R.$$

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$$\begin{split} &(\text{vec}(I_q))^T \left(e_j e_\ell^T \otimes C \right) \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes A - \sum_{i=1}^m \tilde{N}_i \otimes N_i \right)^{-1} \left(\tilde{B} \otimes B \right) \text{vec}(I_m) \\ &= (\text{vec}(I_q))^T \left(e_j e_\ell^T \otimes \hat{C} \right) \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes \hat{A} - \sum_{i=1}^m \tilde{N}_i \otimes \hat{N}_i \right)^{-1} \left(\tilde{B} \otimes \hat{B} \right) \text{vec}(I_m). \end{split}$$



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Using Λ , \tilde{N}_i , \tilde{B} , \tilde{C} as optimization parameters, we can derive necessary conditions for \mathcal{H}_2 -optimality, e.g.:

$$\begin{split} & \left(\text{vec}(\textit{I}_{q}) \right)^{T} \left(e_{j} e_{\ell}^{T} \otimes \textit{C} \right) \left(- \Lambda \otimes \textit{I}_{n} - \textit{I}_{\hat{n}} \otimes \textit{A} - \sum_{i=1}^{m} \tilde{\textit{N}}_{i} \otimes \textit{N}_{i} \right)^{-1} \left(\tilde{\textit{B}} \otimes \textit{B} \right) \text{vec}(\textit{I}_{m}) \\ & = \left(\text{vec}(\textit{I}_{q}) \right)^{T} \left(e_{j} e_{\ell}^{T} \otimes \hat{\textit{C}} \right) \left(- \Lambda \otimes \textit{I}_{n} - \textit{I}_{\hat{n}} \otimes \hat{\textit{A}} - \sum_{i=1}^{m} \tilde{\textit{N}}_{i} \otimes \hat{\textit{N}}_{i} \right)^{-1} \left(\tilde{\textit{B}} \otimes \hat{\textit{B}} \right) \text{vec}(\textit{I}_{m}). \end{split}$$

Where is the connection to the interpolation of transfer functions?

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→ tangential interpolation at mirror images of reduced system poles



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Note: [FLAGG 2011] shows equivalence to interpolating the Volterra series!



Thist iterative Appro

Algorithm 1 Bilinear IRKA

Input: $A, N_i, B, C, \hat{A}, \hat{N}_i, \hat{B}, \hat{C}$

Output: A^{opt} , N_i^{opt} , B^{opt} , C^{opt}

1: **while** (change in $\Lambda > \epsilon$) **do**

2:
$$R\Lambda \hat{R}^{-1} = \hat{A}, \ \tilde{B} = \hat{R}^{-1}\hat{B}, \ \tilde{C} = \hat{C}R, \ \tilde{N}_i = \hat{R}^{-1}\hat{N}_iR$$

3:
$$\operatorname{vec}(V) = \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes A - \sum_{i=1}^m \tilde{N}_i \otimes N_i\right)^{-1} \left(\tilde{B} \otimes B\right) \operatorname{vec}(I_m)$$

4:
$$\operatorname{vec}(W) = \left(-\Lambda \otimes I_n - I_{\hat{n}} \otimes A^T - \sum_{i=1}^m \tilde{N}_i^T \otimes N_i^T\right)^{-1} \left(\tilde{C}^T \otimes C^T\right) \operatorname{vec}(I_q)$$

5:
$$V = \operatorname{orth}(V), W = \operatorname{orth}(W)$$

6:
$$\hat{A} = (W^T V)^{-1} W^T A V, \ \hat{N}_i = (W^T V)^{-1} W^T N_i V, \ \hat{B} = (W^T V)^{-1} W^T B, \ \hat{C} = C V$$

7: end while

8:
$$A^{opt} = \hat{A}$$
, $N_i^{opt} = \hat{N}_i$, $B^{opt} = \hat{B}$, $C^{opt} = \hat{C}$

Numerical Examples



A Heat Transfer Model

- 2-dimensional heat distribution [B./Saak '05]
- Boundary control by spraying intensities of a cooling fluid

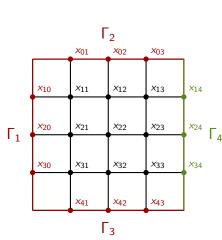
$$\Omega = (0,1) \times (0,1),$$
 $x_t = \Delta x$ in Ω ,
 $n \cdot \nabla x = c \cdot u_{1,2,3}(x-1)$ on $\Gamma_1, \Gamma_2, \Gamma_3$,
 $x = u_4$ on Γ_4 .

• Spatial discretization $k \times k$ -grid

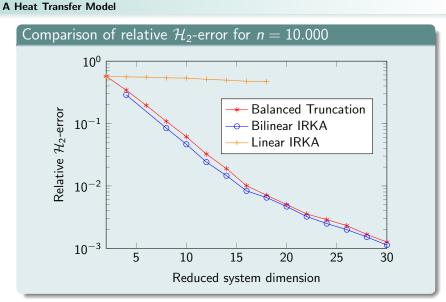
$$\Rightarrow \dot{x} \approx A_1 x + \sum_{i=1}^{3} N_i x u_i + Bu$$

$$\Rightarrow A_2 = 0.$$

• Output: $y = \frac{1}{k^2} \begin{bmatrix} 1 & \dots & 1 \end{bmatrix}$.







Fokker-Planck Equation

As a second example, we consider a dragged Brownian particle whose one-dimensional motion is given by

$$dX_t = -\nabla V(X_t, t)dt + \sqrt{2\sigma}dW_t,$$

with $\sigma = \frac{2}{3}$ and $V(x, u) = W(x, t) + \Phi(x, u_t) = (x^2 - 1)^2 - xu - x$. Alternatively, one can consider ([HARTMANN ET AL. '10]),

$$\rho(x,t)dx = \mathbf{P}\left[X_t \in [x,x+dx)\right]$$

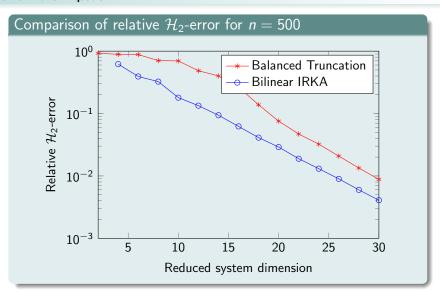
which is described by the Fokker-Planck equation

$$\frac{\partial \rho}{\partial t} = \sigma \Delta \rho + \nabla \cdot (\rho \nabla V), \qquad (x, t) \in (-2, 2) \times (0, T],
0 = \sigma \nabla \rho + \rho \nabla B, \qquad (x, t) \in \{-2, 2\} \times [0, T],
\rho_0 = \rho, \qquad (x, t) \in (-2, 2) \times 0.$$

Output C discrete characteristic function of the interval [0.95, 1.05].

Numerical Examples

Fokker-Planck Equation





Quadratic-Bilinear Differential Algebraic Equations (QBDAEs)

Coming back to the more general case with nonlinear f(x), we consider the class of quadratic-bilinear differential algebraic equations

$$\Sigma: \begin{cases} E\dot{x}(t) = A_1x(t) + A_2x(t) \otimes x(t) + Nx(t)u(t) + Bu(t), \\ y(t) = Cx(t), \quad x(0) = x_0, \end{cases}$$

where $E,A_1,N\in\mathbb{R}^{n\times n},A_2\in\mathbb{R}^{n\times n^2}$ (Hessian tensor), $B,C^T\in\mathbb{R}^n$ are quite helpful.

- A large class of smooth nonlinear control-affine systems can be transformed into the above type of control system.
- The transformation is exact, but a slight increase of the state dimension has to be accepted.



Transformation via McCormick Relaxation

Theorem [Gu'09]

Assume that the state equation of a nonlinear system Σ is given by

$$\dot{x} = a_0 x + a_1 g_1(x) + \ldots + a_k g_k(x) + Bu,$$

where $g_i(x): \mathbb{R}^n \to \mathbb{R}^n$ are compositions of uni-variable rational, exponential, logarithmic, trigonometric or root functions, respectively. Then, by iteratively taking derivatives and adding algebraic equations, respectively, Σ can be transformed into a system of QBDAEs.

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$$E\dot{x}_1 = A_1x_1 + Bu,$$

 $E\dot{x}_2 = A_1x_2 + A_2x_1 \otimes x_1 + Nx_1u,$
 $E\dot{x}_3 = A_1x_3 + A_2(x_1 \otimes x_2 + x_2 \otimes x_1) + Nx_2u$



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$$\begin{split} & \dot{E}\dot{x}_1 = A_1x_1 + Bu, \\ & \dot{E}\dot{x}_2 = A_1x_2 + A_2x_1 \otimes x_1 + Nx_1u, \\ & \dot{E}\dot{x}_3 = A_1x_3 + A_2\left(x_1 \otimes x_2 + x_2 \otimes x_1\right) + Nx_2u \\ & \vdots \end{split}$$

• although i-th subsystem is coupled nonlinearly to preceding systems, linear systems are obtained if terms x_j , j < i, are interpreted as pseudo-inputs.

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$$H_2(s_1, s_2) = \frac{1}{2!}C((s_1 + s_2)E - A_1)^{-1}[N(G_1(s_1) + G_1(s_2)) + A_2(G_1(s_1) \otimes G_1(s_2) + G_1(s_2) \otimes G_1(s_1))],$$



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$$H_{2}(s_{1}, s_{2}) = \frac{1}{2!}C\left((s_{1} + s_{2})E - A_{1}\right)^{-1}\left[N\left(G_{1}(s_{1}) + G_{1}(s_{2})\right) + A_{2}\left(G_{1}(s_{1}) \otimes G_{1}(s_{2}) + G_{1}(s_{2}) \otimes G_{1}(s_{1})\right)\right],$$

$$H_{3}(s_{1}, s_{2}, s_{3}) = \frac{1}{3!}C\left((s_{1} + s_{2} + s_{3})E - A_{1}\right)^{-1}$$

$$\left[N\left(G_{2}(s_{1}, s_{2}) + G_{2}(s_{2}, s_{3}) + G_{2}(s_{1}, s_{3})\right) + A_{2}\left(G_{1}(s_{1}) \otimes G_{2}(s_{2}, s_{3}) + G_{1}(s_{2}) \otimes G_{2}(s_{1}, s_{3}) + G_{1}(s_{2}) \otimes G_{2}(s_{1}, s_{3}) \otimes G_{2}(s_{1}, s_{3}) + G_{2}(s_{2}, s_{3}) \otimes G_{1}(s_{1}) + G_{2}(s_{1}, s_{3}) \otimes G_{1}(s_{2}) + G_{2}(s_{1}, s_{2}) \otimes G_{1}(s_{3})\right)\right].$$



Characterization via Multimoments

For simplicity, focus on the first two transfer functions. For $H_1(s_1)$, choosing σ and making use of the Neumann lemma leads to

$$H_1(s_1) = \sum_{i=0}^{\infty} C \underbrace{\left((A_1 - \sigma E)^{-1} E \right)^i (A_1 - \sigma E)^{-1} B (s_1 - \sigma)^i}_{m_{s_1, \sigma}^i}.$$



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Similarly, specifying an expansion point (au, ξ) yields

$$H_2(s_1, s_2) = \frac{1}{2} \sum_{i=0}^{\infty} C \left((A_1 - (\tau + \xi)E)^{-1}E \right)^i (A_1 - (\tau + \xi)E)^{-1} (s_1 + s_2 - \tau - \xi)^i.$$

$$\left[A_{2}\left(\sum_{j=0}^{\infty}m_{s_{1},\tau}^{j}\otimes\sum_{k=0}^{\infty}m_{s_{2},\xi}^{k}+\sum_{k=0}^{\infty}m_{s_{2},\xi}^{k}\otimes\sum_{j=0}^{\infty}m_{s_{1},\tau}^{j}\right)+N\left(\sum_{p=0}^{\infty}m_{s_{1},\tau}^{p}+\sum_{p=0}^{\infty}m_{s_{2},\xi}^{q}\right)\right]$$



Constructing the Projection Matrix

$$\text{Goal: } \frac{\partial}{\partial s_1^{q-1}} H_1(\sigma) = \frac{\partial}{\partial s_1^{q-1}} \hat{H}_1(\sigma), \quad \frac{\partial}{\partial s_1^{l} s_2^{m}} H_2(\sigma,\sigma) = \frac{\partial}{\partial s_1^{l} s_2^{m}} \hat{H}_2(\sigma,\sigma), \ \ l+m \leq q-1.$$

Construct the following sequence of nested Krylov subspaces



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for $i = 1 : q$

$$V_{2}^{i} = \mathcal{K}_{q-i+1} ((A_{1} - 2\sigma E)^{-1} E, (A_{1} - 2\sigma E)^{-1} NV_{1}(:, i)),$$



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 $V_1(:,i)$ denoting the i-th column of V_1 .



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 $V_1(:,i)$ denoting the i-th column of V_1 . Set $\mathcal{V} = \operatorname{orth} [V_1, V_2^i, V_3^{i,j}]$ and construct $\hat{\Sigma}$ by the Galerkin-Projection $\mathcal{P} = \mathcal{V}\mathcal{V}^T$:

$$\begin{split} \hat{A}_1 &= \mathcal{V}^T A_1 \mathcal{V} \in \mathbb{R}^{\hat{n} \times \hat{n}}, \quad \hat{A}_2 &= \mathcal{V}^T A_2 (\mathcal{V} \otimes \mathcal{V}) \in \mathbb{R}^{\hat{n} \times \hat{n}^2}, \\ \hat{N} &= \mathcal{V}^T N \mathcal{V} \in \mathbb{R}^{\hat{n} \times \hat{n}}, \quad \hat{b} &= \mathcal{V}^T b \in \mathbb{R}^{\hat{n}}, \quad \hat{c}^T = c^T \mathcal{V} \in \mathbb{R}^{\hat{n}}. \end{split}$$



Tensors and Matricizations: A Short Excursion [Kolda/Bader '09, Grasedyck '10]

A tensor is a vector

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indexed by a product index set

$$\mathcal{I} = \mathcal{I}_1 \times \cdots \times \mathcal{I}_d, \quad \# \mathcal{I}_j = \textit{n}_j.$$



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Example: For a given 3-tensor $A_{(i_1,i_2,i_3)}$ with $i_1,i_2,i_3 \in \{1,2\}$, we have:

$$A^{(1)} = \begin{bmatrix} A_{(1,1,1)} & A_{(1,2,1)} & A_{(1,1,2)} & A_{(1,2,2)} \\ A_{(2,1,1)} & A_{(2,2,1)} & A_{(2,1,2)} & A_{(2,2,2)} \end{bmatrix},$$

$$A^{(2)} = \begin{bmatrix} A_{(1,1,1)} & A_{(2,1,1)} & A_{(1,1,2)} & A_{(2,1,2)} \\ A_{(1,2,1)} & A_{(2,2,1)} & A_{(1,2,2)} & A_{(2,2,2)} \end{bmatrix}.$$



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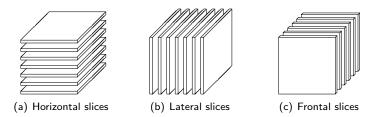


Figure: Slices of a 3rd-order tensor. [Courtesy of Tammy Kolda]



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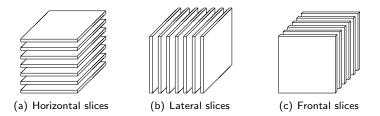


Figure: Slices of a 3rd-order tensor. [Courtesy of Tammy Kolda]

→ Allows to compute matrix products more efficiently.

Two-Sided Projection Methods



Similarly to the linear case, one can exploit duality concepts, in order to construct two-sided projection methods.

Ø

Two-Sided Projection Methods

Similarly to the linear case, one can exploit duality concepts, in order to construct two-sided projection methods.

Interpreting $\mathcal{A}^{(2)}$ now as the 2-matricization of the Hessian 3-tensor corresponding to A_2 , one can show that the dual Krylov spaces have to be constructed as follows

$$\begin{split} W_1 = & \mathcal{K}_q \left((A_1 - 2\sigma E)^{-T} E^T, (A_1 - 2\sigma E)^{-T} c \right) \\ \text{for } i = 1:q \\ W_2^i = & \mathcal{K}_{q-i+1} \left((A_1 - \sigma E)^{-T} E^T, (A_1 - \sigma E)^{-T} N^T W_1(:,i) \right), \\ \text{for } j = 1: \min(q-i+1,i) \\ W_3^{i,j} = & \mathcal{K}_{q-i-j+2} \left((A_1 - \sigma E)^{-T} E^T, (A_1 - \sigma E)^{-T} \mathcal{A}^{(2)} V_1(:,i) \otimes W_1(:,j) \right), \end{split}$$

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Note: Due to the symmetry of the Hessian tensor, the 3-matricization $\mathcal{A}^{(3)}$ coincides with $\mathcal{A}^{(2)}$.



Multimoment matching

Theorem,

- $\Sigma = (E, A_1, A_2, N, b, c)$ original QBDAE system.
- Reduced system by Petrov-Galerkin projection $\mathcal{P} = \mathcal{V}\mathcal{W}^T$ with

$$\begin{split} V_1 &= \mathcal{K}_{q_1} \left(E, A_1, b, \sigma \right), \quad W_1 &= \mathcal{K}_{q_1} \left(E^T, A_1^T, c, 2\sigma \right) \\ \text{for } i &= 1: q_2 \\ V_2 &= \mathcal{K}_{q_2 - i + 1} \left(E, A_1, NV_1(:, i), 2\sigma \right) \\ W_2 &= \mathcal{K}_{q_2 - i + 1} \left(E^T, A_1^T, N^T W_1(:, i), \sigma \right) \\ \text{for } j &= 1: \min(q_2 - i + 1, i) \\ V_3 &= \mathcal{K}_{q_2 - i - j + 2} \left(E, A_1, A_2 V_1(:, i) \otimes V_1(:, j), 2\sigma \right) \\ W_3 &= \mathcal{K}_{q_2 - i - j + 2} \left(E^T, A_1^T, \mathcal{A}^{(2)} V_1(:, i) \otimes W_1(:, j), \sigma \right). \end{split}$$

Then, it holds:

$$\frac{\partial^{i} H_{1}}{\partial s_{1}^{i}}(\sigma) = \frac{\partial^{i} \hat{H}_{1}}{\partial s_{1}^{i}}(\sigma), \quad \frac{\partial^{i} H_{1}}{\partial s_{1}^{i}}(2\sigma) = \frac{\partial^{i} \hat{H}_{1}}{\partial s_{1}^{i}}(2\sigma), \quad i = 0, \dots, q_{1} - 1,$$

$$\frac{\partial^{i+j}}{\partial s_{1}^{i} s_{2}^{j}} H_{2}(\sigma, \sigma) = \frac{\partial^{i+j}}{\partial s_{1}^{i} s_{2}^{j}} \hat{H}_{2}(\sigma, \sigma), \qquad i + j \leq 2q_{2} - 1.$$



Two-Dimensional Burgers Equation

• 2D-Burgers equation on
$$\underbrace{(0,1)\times(0,1)}_{:=\Omega}\times[0,T]$$

$$u_t = -(u \cdot \nabla) u + \nu \Delta u$$

with $u(x, y, t) \in \mathbb{R}^2$ describing the motion of a compressible fluid.



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Consider initial and boundary conditions

$$\begin{split} u_x(x,y,0) &= \frac{\sqrt{2}}{2}, \quad u_y(x,y,0) = \frac{\sqrt{2}}{2}, \qquad \text{for } (x,y) \in \Omega_1 := (0,0.5], \\ u_x(x,y,0) &= 0, \qquad u_y(x,y,0) = 0, \qquad \text{for } (x,y) \in \Omega \backslash \Omega_1, \\ u_x &= 0, \qquad u_y = 0, \qquad \text{for } (x,y) \in \partial \Omega. \end{split}$$

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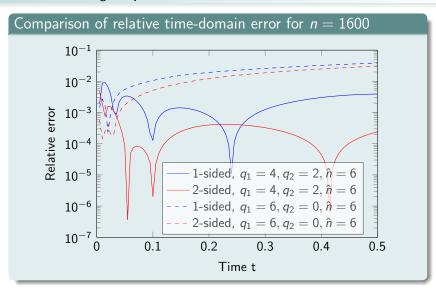
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- Output C chosen to be average x-velocity.

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• Spatial discretization \rightsquigarrow QBDAE system with zero I.C. and 4 inputs $B \in \mathbb{R}^{n \times 4}$, N_1, N_2, N_3, N_4 , ROM with $q_1 = 5, q_2 = 2, \sigma = 0, \hat{n} = 52$.

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- State reconstruction by reduced model $x \approx V\hat{x}$, max. rel. err < 3%.



The Chafee-Infante equation

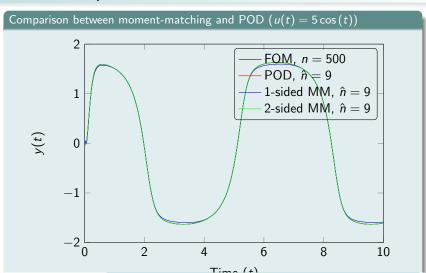
Consider PDE with a cubic nonlinearity:

$$v_t + v^3 = v_{xx} + v,$$
 in $(0,1) \times (0, T),$ $v(0, \cdot) = u(t),$ in $(0, T),$ $v_x(1, \cdot) = 0,$ in $(0, T),$ $v(x, 0) = v_0(x),$ in $(0, T)$

• original state dimension n = 500, QBDAE dimension $N = 2 \cdot 500$, reduced QBDAE dimension r = 9

Nonlinear Model Reduction by Generalized Moment-Matching

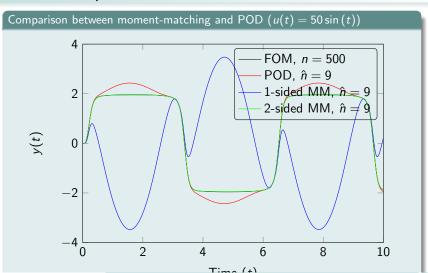
The Chafee-Infante equation



Nonlinear Model Reduction by Generalized Moment-Matching



The Chafee-Infante equation



Nonlinear Model Reduction by Generalized Moment-Matching



The FitzHugh-Nagumo System

• FitzHugh-Nagumo system modeling a neuron

[Chaturantabut, Sorensen '09]

$$\epsilon v_t(x,t) = \epsilon^2 v_{xx}(x,t) + f(v(x,t)) - w(x,t) + g,$$

$$w_t(x,t) = hv(x,t) - \gamma w(x,t) + g,$$

with f(v) = v(v - 0.1)(1 - v) and initial and boundary conditions

$$v(x,0) = 0,$$
 $w(x,0) = 0,$ $x \in [0,1],$ $v_x(0,t) = -i_0(t),$ $v_x(1,t) = 0,$ $t \ge 0,$

where

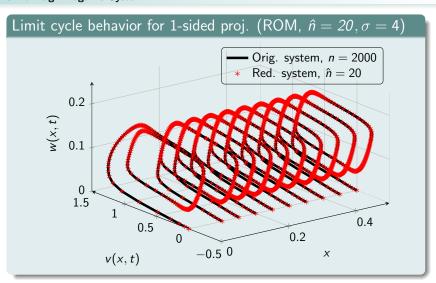
$$\epsilon = 0.015, h = 0.5, \gamma = 2, g = 0.05, i_0(t) = 5 \cdot 10^4 t^3 \exp(-15t)$$

• original state dimension $n = 2 \cdot 1000$, QBDAE dimension $N = 3 \cdot 1000$, reduced QBDAE dimension r = 20



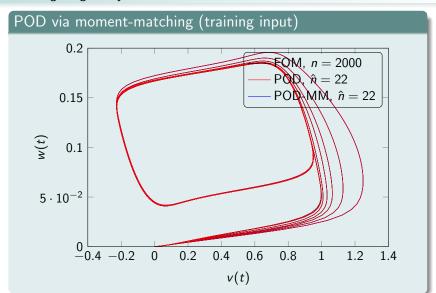
The FitzHugh-Nagumo System





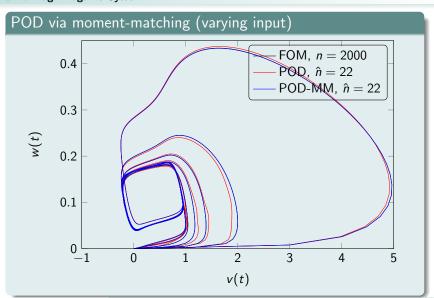
The FitzHugh-Nagumo System







The FitzHugh-Nagumo System





- Many nonlinear dynamics can be expressed by a system of quadratic-bilinear differential algebraic equations.
- For this type of systems, a frequency domain analysis leads to certain generalized transfer functions.
- In contrast to other methods like TPWL and POD, the reduction process is independent of the control input.

Conclusions and Outlook



- Many nonlinear dynamics can be expressed by a system of quadratic-bilinear differential algebraic equations.
- For this type of systems, a frequency domain analysis leads to certain generalized transfer functions.
- In contrast to other methods like TPWL and POD, the reduction process is independent of the control input.
- Optimal choice of interpolation points?
- Stability/index-preserving reduction possible?

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