

MAX PLANCK INSTITUTE FOR DYNAMICS OF COMPLEX TECHNICAL SYSTEMS MAGDEBURG



COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY

Numerical Computation of Robust Controllers for Incompressible Flow Problems Peter Benner, Jan Heiland, Steffen W. R. Werner

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Introduction Flow Control Problem I



Problem: The steady state tends to unstable flow due to unavoidable system perturbations.



Feedback Control

Introduction Flow Control Problem I



Problem: The steady state tends to unstable flow due to unavoidable system perturbations.

Goal: Stabilizing feedback controller that can handle:

- limited measurements,
- short evaluation time,
- system uncertainties.





Idea: Linearization-based feedback control for stabilization of the steady state. [RAYMOND '06, BREITEN/KUNISCH '14, B./HEILAND '15]

$$\dot{v} + (v \cdot \nabla v) - \frac{1}{Re} \Delta v + \nabla p = Bu,$$

 $\nabla \cdot v = 0$
Linearization & $E\dot{v} - Av - J^{T}p = Bu,$
 $Jv = 0$
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$$\nabla \cdot v = 0$$
Linearization &
Semi-Discretization &
 $Jv = 0$

Fragility of Observer-Based Controllers

LQG controllers have no guaranteed robustness margins and will likely fail in the presence of system uncertainties.



Doyle '78



In general, an uncertainty \mathbf{A}_{Δ} in the linearization A ...

$$\dot{x}(t) = Ax(t) + Bu(t),$$

$$y(t) = Cx(t)$$

$$\dot{x}(t) = [A + \mathbf{A}_{\Delta}]x(t) + Bu(t),$$

$$y(t) = Cx(t)$$

... is an additive uncertainty in the transfer function

$$G(s) = C(sI - A)^{-1}B$$

$$G_{\Delta}(s) = C(sI - A - A_{\Delta})^{-1}B$$

= $G(s) + \widetilde{G}(s)$

where $\widetilde{G}(s) = C \mathbf{A}_{\Delta} (sI - A)^{-1} (sI - A - \mathbf{A}_{\Delta})^{-1} B$.



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.

Additive uncertainties can be compensated by robust \mathcal{H}_{∞} controller design.



Robust Controller Design

Linear Time-Invariant Systems

$$G:\begin{cases} \dot{x} = Ax + B_1w + B_2u, \\ z = C_1x + D_{11}w + D_{12}u, \\ y = C_2x + D_{21}w + D_{22}u, \end{cases}$$

where A, B_j, C_i, D_{ij} are matrices of suitable size, $j, i \in \{1, 2\}$.





For such partitioned systems, the transfer function writes

$$G(s) = \begin{bmatrix} G_{11}(s) & G_{12}(s) \\ G_{21}(s) & G_{22}(s) \end{bmatrix} = \begin{bmatrix} A & B_1 & B_2 \\ \hline C_1 & D_{11} & D_{12} \\ C_2 & D_{21} & D_{22} \end{bmatrix},$$

such that

$$z(s) = G_{11}(s)w(s) + G_{12}u(s),$$

$$y(s) = G_{21}(s)w(s) + G_{22}u(s).$$

Goal: Optimal \mathcal{H}_{∞} Control Problem

Find a controller K that minimizes the performance outputs of the closed-loop system

$$z = (G_{11} + G_{12}K(I - G_{22}K)^{-1}G_{21})w =: \mathcal{F}(G, K)w$$

and internally stabilizes the plant, i.e.,

$$\min_{K \text{ stabilizing}} \|\mathcal{F}(G,K)\|_{\mathcal{H}_{\infty}}.$$



For such partitioned systems, the transfer function writes

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such that

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Goal: Sub-Optimal \mathcal{H}_{∞} Control Problem

Find a controller K that minimizes the performance outputs of the closed-loop system

$$z = (G_{11} + G_{12}K(I - G_{22}K)^{-1}G_{21})w =: \mathcal{F}(G, K)w$$

and internally stabilizes the plant, i.e., for a given constant γ

$$\|\mathcal{F}(G,K)\|_{\mathcal{H}_{\infty}} < \gamma .$$



Robust Controller Design Solution to the \mathcal{H}_{∞} Control Problem

Simplifying Assumptions:

- 1 $D_{11} = 0$
- **2** $D_{22} = 0$
- (A, B_1) stabilizable and (A, C_1) detectable
- **4** (A, B_2) stabilizable and (A, C_2) detectable

$$\mathbf{O} \ D_{12}^{\mathsf{T}} \begin{bmatrix} C_1 & D_{12} \end{bmatrix} = \begin{bmatrix} \mathbf{0} & I \end{bmatrix}$$

$$\mathbf{\mathfrak{o}} \begin{bmatrix} B_1 \\ D_{21} \end{bmatrix} D_{21}^\mathsf{T} = \begin{bmatrix} \mathbf{0} \\ I \end{bmatrix}$$

Remarks:

- Assumptions 2., 5., 6. are for notational convenience.
- From 4., internal stabilizability follows.
- The points 1., 3. can be relaxed, but derivations become more complicated.



\mathcal{H}_{∞} Riccati Equations

[Doyle/Glover/Khargonekar/Francis '89, Van Keulen '93]

Given the assumptions 1.–6., there exists an admissible controller $K(s) \iff$:

(1) There exists a stabilizing solution $X_{\infty} = X_{\infty}^{\mathsf{T}} \ge 0$ to the regulator Riccati equation

$$\boldsymbol{A}^{\mathsf{T}}\boldsymbol{X}_{\infty} + \boldsymbol{X}_{\infty}\boldsymbol{A} + \boldsymbol{C}_{1}^{\mathsf{T}}\boldsymbol{C}_{1} + \boldsymbol{X}_{\infty}(\boldsymbol{\gamma}^{-2}\boldsymbol{B}_{1}\boldsymbol{B}_{1}^{\mathsf{T}} - \boldsymbol{B}_{2}\boldsymbol{B}_{2}^{\mathsf{T}})\boldsymbol{X}_{\infty} = \boldsymbol{0}.$$

2 There exists a stabilizing solution $Y_\infty = Y_\infty^\mathsf{T} \ge 0$ to the filter Riccati equation

$$AY_{\infty} + Y_{\infty}A^{\mathsf{T}} + B_1B_1^{\mathsf{T}} + Y_{\infty}(\gamma^{-2}C_1^{\mathsf{T}}C_1 - C_2^{\mathsf{T}}C_2)Y_{\infty} = 0.$$

3 It holds $\gamma^2 > \lambda_{\max}(Y_{\infty}X_{\infty})$.

The central (or minimum entropy) controller $\widehat{K}(s) = \widehat{C}(sI_n - \widehat{A})^{-1}\widehat{B}$ is given by $\widehat{A} = A + (\gamma^{-2}B_1B_1^{\mathsf{T}} - B_2B_2^{\mathsf{T}})X_{\infty} - Z_{\infty}Y_{\infty}C_2^{\mathsf{T}}C_2, \quad \widehat{B} = Z_{\infty}Y_{\infty}C_2^{\mathsf{T}}, \quad \widehat{C} = -B_2^{\mathsf{T}}X_{\infty},$ with $Z_{\infty} = (I_n - \gamma^{-2}X_{\infty}Y_{\infty})^{-1}$.



Large-Scale DAE Matrix Equations

How to solve the arising large-scale sparse DAE Riccati equations?

• Use low-rank Riccati iteration for approximation of the solution.

[LANZON/FENG/ANDERSON '07, B./HEILAND/WERNER '19]

• Apply **implicit projections** to handle DAE structure.

[Heinkenschloss/Sorensen/Sun '08, Bänsch/B./Saak/Weichelt '15]



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High Dimensional Robust Controller

How much uncertainty can be compensated by a low-dimensional robust controller?

- Robustness margin gives condition on coprime factor perturbations.
- Use model order reduction based on X_{∞} and Y_{∞} . $\Rightarrow \mathcal{H}_{\infty}$ Balanced Truncation



Given a transfer function G(s) of a linear system,

 $G(s) = M^{-1}(s)N(s)$

is a (left) coprime factorization if there exist U(s) and V(s) such that

 $N(s)U(s) + M(s)V(s) = I_{\rho},$

where M, N, U, V are transfer functions with all poles in the open left complex half-plane, i.e., corresponding to stable linear systems.



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Coprime Factor Perturbations

The transfer functions G(s) and $G_{\Delta}(s)$ differ by a coprime factor disturbance if

$$G(s) = M(s)^{-1}N(s)$$
 and $G_{\Delta}(s) = M_{\Delta}(s)^{-1}N_{\Delta}(s),$

with $\Delta = [N - N_{\Delta}, M - M_{\Delta}]$ stable.

Uncertainties in the linearization result in coprime factor perturbations.



Assumption: Normalized LQG problem, i.e., $B_1 = B_2 = B$, $C_1 = C_2 = C$.

Theorem

[McFarlane/Glover '90]

Let the controller K stabilize the plant G with

$$\left[\begin{bmatrix} K \\ I_p \end{bmatrix} (I_p - GK)^{-1} M^{-1} \right]_{\mathcal{H}_{\infty}} < \gamma^{-1},$$

then K stabilizes G_{Δ} if the perturbation Δ is small, namely $\|\Delta\|_{\mathcal{H}_{\infty}} < \gamma$.



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then K stabilizes G_{Δ} if the perturbation Δ is small, namely $\|\Delta\|_{\mathcal{H}_{\infty}} < \gamma$.

Let L be a gain such that

A - LC and $A + A_{\Delta} - LC$ are both stable,

then realizations of the coprime factorizations $G(s) = M(s)^{-1}N(s)$ and $G_{\Delta}(s) = M_{\Delta}(s)^{-1}N_{\Delta}(s)$ are given by

$$\begin{bmatrix} N & M \end{bmatrix} = \begin{bmatrix} A - LC & B & -L \\ \hline C & 0 & I_p \end{bmatrix}, \quad \begin{bmatrix} N_{\Delta} & M_{\Delta} \end{bmatrix} = \begin{bmatrix} A + A_{\Delta} - LC & B & -L \\ \hline C & 0 & I_p \end{bmatrix}$$



Motivation

In practice, the number of differential equations becomes very large.

 \Rightarrow Large demand of computational resources, i.e., time and memory.



Goal

Approximation of the input-output behavior, i.e., $||y - y_r|| \le tol \cdot ||u||$ for all admissible input signals u.



Balancing Related Model Reduction Methods

- \bullet Solve primal and dual matrix equations defining the characteristic matrices P and Q.
- **2** Balance the system with respect to P and Q.
- 3 Truncate states corresponding to small characteristic values of PQ.



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- **2** Balance the system with respect to P and Q.
- 3 Truncate states corresponding to small characteristic values of PQ.
- Linear-Quadratic Gaussian Balanced Truncation (LQGBT):

[JONCKHEERE/SILVERMAN '83]

$$A\boldsymbol{P}_{LQG} + \boldsymbol{P}_{LQG}A^{\mathsf{T}} - \boldsymbol{P}_{LQG}C^{\mathsf{T}}C\boldsymbol{P}_{LQG} + BB^{\mathsf{T}} = 0,$$

$$A^{\mathsf{T}}\boldsymbol{Q}_{LQG} + \boldsymbol{Q}_{LQG}A - \boldsymbol{Q}_{LQG}BB^{\mathsf{T}}\boldsymbol{Q}_{LQG} + C^{\mathsf{T}}C = 0.$$

• \mathcal{H}_{∞} Balanced Truncation (HINFBT): [MUSTAFA/GLOVER '91]

$$A\boldsymbol{P}_{\mathcal{H}_{\infty}} + \boldsymbol{P}_{\mathcal{H}_{\infty}}A^{\mathsf{T}} + \boldsymbol{P}_{\mathcal{H}_{\infty}}(\gamma^{-2}C^{\mathsf{T}}C - C^{\mathsf{T}}C)\boldsymbol{P}_{\mathcal{H}_{\infty}} + BB^{\mathsf{T}} = 0,$$

$$A^{\mathsf{T}}\boldsymbol{Q}_{\mathcal{H}_{\infty}} + \boldsymbol{Q}_{\mathcal{H}_{\infty}}A + \boldsymbol{Q}_{\mathcal{H}_{\infty}}(\gamma^{-2}BB^{\mathsf{T}} - BB^{\mathsf{T}})\boldsymbol{Q}_{\mathcal{H}_{\infty}} + C^{\mathsf{T}}C = 0.$$

Assumptions/Notations:

CSC

- normalized LQG problem, i.e., $B_1 = B_2 = B$, $C_1 = C_2 = C$,
- normalized left coprime factorizations G = M⁻¹N, G_r = M⁻¹_rN_r, G_Δ = M⁻¹_ΔN_Δ (for construction see [B./HEILAND/WERNER '19]),
- $\beta = \sqrt{1 \gamma^{-2}}.$

The approximation error of the \mathcal{H}_∞ balanced truncation is given by

$$\left\| \begin{bmatrix} \beta(N-N_r) & M-M_r \end{bmatrix} \right\|_{\mathcal{H}_{\infty}} \leq \beta \epsilon = 2 \sum_{k=r+1}^n \frac{\beta \sigma_k^{\mathcal{H}_{\infty}}}{\sqrt{1+\beta^2 \left(\sigma_k^{\mathcal{H}_{\infty}}\right)^2}}.$$

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[Mustafa/Glover '91, B./Heiland/Werner '19]

The reduced-order \mathcal{H}_∞ controller is guaranteed to stabilize the disturbed system if

$$\epsilon(eta+\gamma) < 1$$
 and $\left\| \begin{bmatrix} N - N_{\Delta} & M - M_{\Delta} \end{bmatrix} \right\|_{\mathcal{H}_{\infty}} < \gamma^{-1}.$

Theorem





Fig.: 2D cylinder wake, discretized by Taylor-Hood (P_2/P_1) finite elements.

- Navier-Stokes equations
- Reynolds number 90
- 9,843 velocity nodes
- distributed observations:
 - 3 sensors in the wake
 - measuring both v components

Numerical Example

Target

Stabilize the steady-state and compensate perturbations to suppress vortex shedding.

- boundary control:
 - 2 outlets at the cylinder periphery
 - control by injection and suction •





- error in linearization: 8%
- reduced-order controller dimension: 7
- trigger instabilities by input disturbance on time interval [0, 1]:

$$u_{\delta}(t) = egin{bmatrix} 0.01\sin(2t\pi)\ -0.01\sin(2t\pi) \end{bmatrix}$$





$$u_{\delta}(t) = egin{bmatrix} 0.01\sin(2t\pi) \ -0.01\sin(2t\pi) \end{bmatrix}$$



Summary

 $\mathcal{H}_\infty\text{-}\mathsf{BT}$ controllers are

- output-based and of low dimension,
- robust against system uncertainties.

The application to flow stabilization becomes feasible by

- the implicit realization of the DAE structure,
- the low-rank Riccati iteration.



Summary

 $\mathcal{H}_\infty\text{-}\mathsf{BT}$ controllers are

- output-based and of low dimension,
- robust against system uncertainties.

The application to flow stabilization becomes feasible by

- the implicit realization of the DAE structure,
- the low-rank Riccati iteration.

Code Availability:

- The low-rank Riccati iteration will be available in the M-M.E.S.S. library version 2.0.
- HINFBT and LQGBT implementations can be found in the MORLAB toolbox.







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