

MAX PLANCK INSTITUTE FOR DYNAMICS OF COMPLEX TECHNICAL SYSTEMS MAGDEBURG



COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY

Balanced Truncation for Linear and Nonlinear Systems on Industrial Scale

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1. Introduction

Model Order Reduction of Dynamical Systems Industrial Application Areas System Classes Model Reduction of Linear Systems Model Reduction in Frequency Domain

- 2. Balanced Truncation for Linear Systems
- 3. Balanced Truncation for Polynomial Systems
- 4. Conclusions



Original System

$$\Sigma : \begin{cases} \dot{x}(t) &= f(t, x(t), u(t)), \\ y(t) &= g(t, x(t), u(t)), \end{cases}$$

- states $x(t) \in \mathbb{R}^n$,
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Introduction Model Order Reduction of Dynamical Systems

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Reduced-Order Model (ROM)

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$$\xrightarrow{u}$$
 $\widehat{\Sigma}$ $\xrightarrow{\widehat{y}}$

Goals:

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Feedback Control

A feedback controller (dynamic compensator) is a linear system of order N, where

- input = output of plant,
- output = input of plant.

Modern (LQG-/ \mathcal{H}_2 -/ \mathcal{H}_∞ -) control design: $N \ge n$.



Sources: MPI Magdeburg (pendulum), https://tinyurl.com/9n75h5dd under CC BY 2.0 license (drone).



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Practical controllers require small N ($N \sim 10$, say) due to

- real-time constraints,
- increasing fragility for larger N.



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```
\implies reduce order of plant (n) and/or controller (N).
```



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• Progressive miniaturization: Moore's Law states that the number of on-chip transistors doubles each 12 (now: 18) months.



Combline Diplexer: n = 270, 446, [Zhao/Wu, ieee-mtt 2014]



Substrate Integrated Waveguide Antenna: n = 390, 302, [Dong/Itoh, ieee-tap 2010]



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- Linear systems in nanoelectronics occur through
 - modified nodal analysis (MNA) for RLC networks, e.g., when decoupling large linear subcircuits,
 - modeling transmission lines (interconnect, powergrid), parasitic effects,
 - modeling circuit elements described by Maxwell's equation using partial element equivalent circuits (PEEC), e.g., microwave devices like splitters and diplexers, electromagnetic devices like antennas.



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- Resolving complex 3D geometries \Rightarrow millions of degrees of freedom.
- Analysis of elastic deformations requires many simulation runs for varying external forces.
- Additional goal: Preserve second-order structure in reduced-order model, i.e.,

$$M\ddot{x}(t) + D\dot{x}(t) + \mathcal{K}(x(t)) = Bu(t)$$

for seamless integration into commercial EMBD simulation software.



$\dot{x}(t)$	=	f(t, x, u)	=	$\mathcal{A}(x(t)) + \mathcal{B}(x(t))u(t),$	$\mathcal{A}: \mathbb{R}^n \to \mathbb{R}^n, \ \mathcal{B}: \mathbb{R}^n \to \mathbb{R}^{n \times m},$
y(t)	=	g(t, x, u)	=	$\mathcal{C}(x(t)) + \mathcal{D}(x(t))u(t),$	$\mathcal{C}: \mathbb{R}^n \to \mathbb{R}^q, \ \mathcal{D}: \mathbb{R}^n \to \mathbb{R}^{q \times m}.$



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Linear Time-Invariant (LTI) Systems

$\dot{x}(t)$	=	f(t, x, u)	=	Ax(t) + Bu(t),	$A \in \mathbb{R}^{n \times n},$	$B \in \mathbb{R}^{n \times m},$
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Bilinear Systems

$\dot{x}(t)$ =	= f(t, x, u) =	$= Ax(t) + \sum_{i=1}^{m} u_i(t) A_i x(t) + Bu(t),$	$A, A_i \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m},$
y(t) =	= g(t, x, u) =	Cx(t) + Du(t),	$C \in \mathbb{R}^{q \times n}, \ D \in \mathbb{R}^{q \times m}.$



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Quadratic-Bilinear (QB) Systems

 $\begin{aligned} \dot{x}(t) &= f(t, x, u) = Ax(t) + H\left(x(t) \otimes x(t)\right) + \sum_{i=1}^{m} u_i(t)A_ix(t) + Bu(t), \\ A, A_i \in \mathbb{R}^{n \times n}, \ H \in \mathbb{R}^{n \times n^2}, \ B \in \mathbb{R}^{n \times m}, \\ y(t) &= g(t, x, u) = Cx(t) + Du(t), \qquad C \in \mathbb{R}^{q \times n}, \ D \in \mathbb{R}^{q \times m}. \end{aligned}$

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Polynomial Systems

$$\begin{aligned} \dot{x}(t) &= f(t, x, u) = Ax(t) + \sum_{j=2}^{n_p} H_j\left(\otimes^j x(t)\right) + \sum_{j=2}^{n_p} \sum_{k=1}^m A_j^k\left(\otimes^j x(t)\right) u_k(t) + Bu(t), \\ H_j, A_j^k \text{ of "appropriate size"}, \end{aligned}$$
$$y(t) &= g(t, x, u) = Cx(t) + Du(t), \qquad C \in \mathbb{R}^{q \times n}, \ D \in \mathbb{R}^{q \times m}. \end{aligned}$$

Polynomial-bilinear systems can be obtained from smooth nonlinear systems by *lifting* without approximation error! [Gu 2011].



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Written in control-affine form:

 $\begin{aligned} \mathcal{A}(x) &:= Ax + H(x \otimes x), \qquad \mathcal{B}(x) &:= [A_1, \dots, A_m](I_m \otimes x) + B \\ \mathcal{C}(x) &:= Cx, \qquad \qquad \mathcal{D}(x) &:= D. \end{aligned}$



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Here, we focus on linear and polynomial systems.



Model Reduction of Linear Systems

Linear Time-Invariant (LTI) Systems

Original System

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Goals:

 $||y - \hat{y}|| < \text{tolerance} \cdot ||u||$ for all admissible input signals. Secondary goal: reconstruct approximation of x from \hat{x} .



Linear Systems in Frequency Domain

Application of Laplace transform $(x(t) \mapsto x(s), \dot{x}(t) \mapsto sX(s) - x(0))$ to LTI system

$$\dot{x}(t) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t)$$

with x(0) = 0 yields:

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 \implies I/O-relation in frequency domain:

$$Y(s) = \left(\underbrace{C(sI_n - A)^{-1}B + D}_{=:G(s)}\right)U(s).$$

G(s) is the **transfer function** of Σ .



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Model reduction in frequency domain: Fast evaluation of mapping $U \rightarrow Y$.



Formulating model reduction in frequency domain

Approximate the time domain dynamical system

$$\begin{split} \dot{x} &= Ax + Bu, \qquad A \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m}, \\ y &= Cx + Du, \qquad C \in \mathbb{R}^{p \times n}, \ D \in \mathbb{R}^{p \times m}, \end{split}$$

by reduced-order system

$$\begin{split} \dot{\hat{x}} &=& \hat{A}\hat{x} + \hat{B}u, \quad \hat{A} \in \mathbb{R}^{r \times r}, \ \hat{B} \in \mathbb{R}^{r \times m}, \\ \hat{y} &=& \hat{C}\hat{x} + \hat{D}u, \quad \hat{C} \in \mathbb{R}^{p \times r}, \ \hat{D} \in \mathbb{R}^{p \times m} \end{split}$$

of order $r \ll n$, such that

$$\begin{split} \|y - \hat{y}\| \simeq \left\| Y - \hat{Y} \right\| &= \left\| GU - \hat{G}U \right\| \\ &\leq \left\| G - \hat{G} \right\| \cdot \|U\| \simeq \left\| G - \hat{G} \right\| \cdot \|u\| \\ &\leq \mathsf{tolerance} \cdot \|u\| \,. \end{split}$$



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• System
$$\Sigma$$
:

$$\begin{cases}
\dot{x}(t) = Ax(t) + Bu(t), \\
y(t) = Cx(t),
\end{cases}$$

with A stable, i.e., $\Lambda\left(A\right)\subset\mathbb{C}^{-}$,

is balanced, if system Gramians, i.e., solutions ${\cal P},{\cal Q}$ of the Lyapunov equations

$$AP + PA^{T} + BB^{T} = 0, \qquad A^{T}Q + QA + C^{T}C = 0,$$

satisfy: $P = Q = \operatorname{diag}(\sigma_1, \ldots, \sigma_n)$ with $\sigma_1 \ge \sigma_2 \ge \ldots \ge \sigma_n > 0$.



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- \bullet Compute balanced realization (needs P,Q!) of the system via state-space transformation

$$\begin{array}{rcl} & & & \quad \ (A,B,C) & \mapsto & (TAT^{-1},TB,CT^{-1}) \\ & & = & \left(\left[\begin{array}{cc} A_{11} & A_{12} \\ A_{21} & A_{22} \end{array} \right], \left[\begin{array}{cc} B_1 \\ B_2 \end{array} \right], \left[\begin{array}{cc} C_1 & C_2 \end{array} \right] \right). \end{array}$$



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$$\begin{aligned} \mathcal{T}: (A, B, C) &\mapsto (TAT^{-1}, TB, CT^{-1}) \\ &= \left(\begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix}, \begin{bmatrix} B_1 \\ B_2 \end{bmatrix}, \begin{bmatrix} C_1 & C_2 \end{bmatrix} \right). \end{aligned}$$

• Truncation $\rightsquigarrow (\hat{A}, \hat{B}, \hat{C}) = (A_{11}, B_1, C_1).$ Note: in efficient algorithms, truncation is achieved via projection:

$$(\hat{A}, \hat{B}, \hat{C}) = (W^T A V, W^T B, C V), \quad \text{where} \quad W^T V = I_r.$$



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- Adaptive choice of r via computable error bound:

$$||y - \hat{y}||_2 \le ||G - \hat{G}||_{\mathcal{H}_{\infty}} ||u||_2 \le \left(2\sum_{k=r+1}^n \sigma_k\right) ||u||_2,$$

where $\|G\|_{\mathcal{H}_{\infty}} := \sup_{u \in \mathcal{L}_2 \setminus \{0\}} \frac{\|Gu\|_2}{\|u\|_2} = \sup_{\omega \in \mathbb{R}} \sigma_{\max}(G(j\omega)).$



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Practical implementation

- Rather than solving Lyapunov equations for P, Q (n^2 unknowns!), find $S, R \in \mathbb{R}^{n \times s}$ with $s \ll n$ such that $P \approx SS^T$, $Q \approx RR^T$.
- Reduced-order model directly obtained via small-scale ($s \times s$) SVD of $R^T S!$

^ahttp://www.mpi-magdeburg.mpg.de/projects/morlab

^bhttps://www.mpi-magdeburg.mpg.de/projects/mess, full MATLAB integration in progress.



- Reduced-order model is stable with HSVs $\sigma_1, \ldots, \sigma_r$.
- Adaptive choice of r via computable error bound:

$$|y - \hat{y}||_2 \le ||G - \hat{G}||_{\mathcal{H}_{\infty}} ||u||_2 \le \left(2\sum_{k=r+1}^n \sigma_k\right) ||u||_2,$$

where $||G||_{\mathcal{H}_{\infty}} := \sup_{u \in \mathcal{L}_2 \setminus \{0\}} \frac{||Gu||_2}{||u||_2} = \sup_{\omega \in \mathbb{R}} \sigma_{\max}(G(j\omega)).$

Practical implementation

- Rather than solving Lyapunov equations for P, Q (n^2 unknowns!), find $S, R \in \mathbb{R}^{n \times s}$ with $s \ll n$ such that $P \approx SS^T$, $Q \approx RR^T$.
- Reduced-order model directly obtained via small-scale $(s \times s)$ SVD of $R^T S!$
- Two software packages:
 - MORLAB^a (Model Order Reduction LABoratory), based on spectral projection methods (\rightsquigarrow small to medium size problems, up to $n \sim 5,000.$)
 - M-M.E.S.S.^b provides solvers for large-scale matrix equations with sparse/low-rank coefficients and basic MOR functionality; no $\mathcal{O}(n^3)$ or $\mathcal{O}(n^2)$ computations necessary!

^ahttp://www.mpi-magdeburg.mpg.de/projects/morlab

^bhttps://www.mpi-magdeburg.mpg.de/projects/mess, full MATLAB integration in progress.



• $\operatorname{SIMPLORER}^{\textcircled{R}}$ test circuit with 2 transistors.



- Conservative thermic sub-system in SIMPLORER: voltage ~→ temperature, current ~→ heat flow.
- Original model: n = 270.593, m = q = 2 ⇒
 Computing times (on Intel Xeon dualcore 3GHz, 1 Thread):
 - Main computational cost for set-up data $\approx 22min$.
 - Computation of reduced models from set-up data: 44–49sec. (r = 20-70).
 - Bode plot (MATLAB on Intel Core i7, 2,67GHz, 12GB):
 7.5h for original system, < 1min for reduced system.


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[Siebelts/Saak/Werner '19]

Original system:

- mechanical second-order structure
- n = 779, 232, m = 1, p = 3
- test simulation time $\approx 4\,\mathrm{h}$







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- mechanical second-order structure
- n = 779, 232, m = 1, p = 3
- test simulation time $\approx 4\,\mathrm{h}$

Second-Order Balanced Truncation:

- two-step method with MORLAB backend
- r = 1, wall time $\approx 21 \text{ h}$
- test simulation time ≈ 10 msec (speed up $\approx 1.4 \cdot 10^6$)









Industrial challenges for virtual twin:

- non-homogeneous initial conditions (IC) two approaches: augment input with IC ("BTX0") or use superposition ("2phase"),
- subsystem reduction ("output coupled") vs. holistic reduction ("FE-coupled").





FE-coupled

output-coupled

method	red. order tol 10^{-3}	t_{red}
2phase	196	6.5h
BTX0	174	4.5h

method	red. order tol 10^{-3}	t_{red}
2phase	3,005	2h
BTX0	2,515	1.8h



FE-coupled

output-coupled

method	red. order tol 10^{-3}	t_{red}	
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method	red. order tol 10^{-3}	t_{red}
2phase	3,005	2h
BTX0	2,515	1.8h

 \rightarrow Required storage for reduced matrices just 1MB!





Vettermann, J., Sauerzapf, S., Naumann, A., Beitelschmidt, M., Herzog, R., Benner, P., Saak, J. (2021): Model order reduction methods for coupled machine tool models. MM Science Journal 2021:4652-4659.





Vortex shedding

• Often encountered in industrial applications: system of the form

 $E\dot{x}(t) = Ax(t) + Bu(t)$

with singular $E \rightarrow$ linear differential-algebraic (descriptor) system

• Here: linearized Navier-Stokes equations for control design: index = 2, n = 22,385, m = 2, p = 7.

Balanced truncation approximation

- r = 153 for tolerance $\tau = 10^{-5}$
- hardware: Intel[®]Core[™] i7-6700 with 16GB RAM
- software: M-M.E.S.S.-2.0.1 in MATLAB[®] R2018a
- wall time: \approx 2min.

M.E.S.S. demo example: bt_mor_DAE2('NSE',3,500)









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step response (FOM solid, ROM dashed)



1. Introduction

- 2. Balanced Truncation for Linear Systems
- 3. Balanced Truncation for Polynomial Systems Balanced Truncation for Nonlinear Systems Polynomial Control Systems Gramians for PC Systems Truncated Gramians Numerical Example

4. Conclusions



• Nonlinear balancing based on energy functionals [SCHERPEN 1993, GRAY/MESKO 1996].

Definition

[Scherpen 1993, Gray/Mesko 1996]

The reachability energy functional $L_c(x_0)$, and observability energy functional $L_o(x_0)$ of a system are given as:

$$L_{c}(x_{0}) = \inf_{\substack{u \in L_{2}(-\infty,0]\\x(-\infty)=0, \ x(0)=x_{0}}} \frac{1}{2} \int_{-\infty}^{0} \|u(t)\|^{2} dt, \qquad L_{o}(x_{0}) = \frac{1}{2} \int_{0}^{\infty} \|y(t)\|^{2} dt.$$

Disadvantage: energy functionals are the solutions of nonlinear Hamilton-Jacobi equations which are hardly solvable for large-scale systems.

Note: For linear (LTI) systems,

$$L_c(x_0) = \frac{1}{2} x_0^T P^{-1} x_0, \qquad \qquad L_o(x_0) = \frac{1}{2} x_0^T Q x_0,$$

where ${\cal P},{\cal Q}$ are the controllability and observability Gramians, respectively!



• Nonlinear balancing based on energy functionals [SCHERPEN 1993, GRAY/MESKO 1996].

Disadvantage: energy functionals are the solutions of nonlinear Hamilton-Jacobi equations which are hardly solvable for large-scale systems.

• Empirical Gramians/frequency-domain POD [Lall et al 1999, Willcox/Peraire 2002].

Example: controllability Gramian from time domain data (snapshots)

1 Define reachability Gramian of the system

$$P = \int_0^\infty x(t)x(t)^T dt$$
, where $x(t)$ solves $\dot{x} = f(x, \delta), \ x(0) = x_0.$

- **2** Use time-domain integrator to produce snapshots $x_k \approx x(t_k)$, $k = 1, \ldots, K$.
- 3 Approximate $P \approx \sum_{k=0}^{K} w_k x_k x_k^T$ with positive weights w_k .
- 4 Analogously for observability Gramian.
- 6 Compute balancing transformation and apply it to nonlinear system.

Disadvantage: Depends on chosen training input (e.g., $\delta(t_0)$) like other POD approaches. For recent developments on empirical Gramians, see [HIMPE 2018].



• Nonlinear balancing based on energy functionals [SCHERPEN 1993, GRAY/MESKO 1996].

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- Empirical Gramians/frequency-domain POD [Lall et al 1999, WILLCOX/PERAIRE 2002]. **Disadvantage:** Depends on chosen training input (e.g., $\delta(t_0)$) like other POD approaches. For recent developments on empirical Gramians, see [HIMPE 2018].
- ~ Goal: computationally efficient and input-independent method!

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• A possible solution is to obtain bounds for the energy functionals, instead of computing them exactly.



- A possible solution is to obtain bounds for the energy functionals, instead of computing them exactly.
- For bilinear systems, such local bounds were derived in [B./DAMM 2011] using the solutions to the Lyapunov-plus-positive equations:

 $AP + PA^{T} + \sum_{i=1}^{m} A_{i}PA_{i}^{T} + BB^{T} = 0,$ $A^{T}Q + QA^{T} + \sum_{i=1}^{m} A_{i}^{T}QA_{i} + C^{T}C = 0.$

(Note: if their solutions exist, they define reachability and observability Gramians of BIBO stable bilinear system.)

- Here, we aim at determining algebraic Gramians for polynomial systems, which
 - provide bounds for the energy functionals of polynomial systems,
 - generalize the Gramians of linear and bilinear systems, and
 - allow us to find the states that are hard to reach as well as hard to observe in an efficient and reliable way.



Now, consider the class of polynomial control (PC) Systems:

$$\dot{x}(t) = Ax(t) + \sum_{j=2}^{n_p} H_j\left(\otimes^j x(t)\right) + \sum_{j=2}^{n_p} \sum_{k=1}^m N_j^k\left(\otimes^j x(t)\right) u_k(t) + Bu(t),$$

$$y(t) = Cx(t), \quad x(0) = 0,$$

where

• n_p is the degree of the polynomial part of the system,

•
$$x(t) \in \mathbb{R}^n$$
, $\otimes^j x(t) = x(t) \otimes \cdots \otimes x(t)$,

j-times

•
$$u(t) \in \mathbb{R}^m$$
, and $y(t) \in \mathbb{R}^p$, $n \gg m, p$.

- $A \in \mathbb{R}^{n \times n}$, $H_j, N_j^k \in \mathbb{R}^{n \times n^j}$, $B \in \mathbb{R}^{n \times m}$ and $C \in \mathbb{R}^{p \times n}$.
- Assumption: A is supposed to be Hurwitz \Rightarrow local stability.



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- Assumption: A is supposed to be Hurwitz \Rightarrow local stability.

Examples: FitzHugh-Nagumo and Chafee-Infante equations lead to cubic control systems; cubic-quintic Allen-Cahn equation to quintic control system.



• Consider input \rightarrow state map of (for simplicity) quadratic-bilinear system ($n_P = 2$, $m = 1, N \equiv A_1$):

$$\dot{x}(t) = Ax(t) + Hx(t) \otimes x(t) + Nx(t)u(t) + Bu(t), \qquad x(0) = 0.$$

Integration yields

$$\begin{aligned} x(t) &= \int\limits_{0}^{t} e^{A\sigma_1} Bu(t-\sigma_1) d\sigma_1 + \int\limits_{0}^{t} e^{A\sigma_1} Nx(t-\sigma_1)u(t-\sigma_1) d\sigma_1 \\ &+ \int\limits_{0}^{t} e^{A\sigma_1} Hx(t-\sigma_1) \otimes x(t-\sigma_1) d\sigma_1 \end{aligned}$$

[Rugh 1981]



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[Rugh 1981]



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 By iteratively inserting expressions for x(t − •), we obtain the Volterra series expansion for quadratic-bilinear (and, more general, polynomial) systems. [RUGH 1981]



Expanding the response of the PC system into a Volterra series representation and using the idea of iterated linear systems, we define the reachability Gramian as

$$P = \sum_{k=1}^{\infty} \int_0^{\infty} \cdots \int_0^{\infty} \bar{P}_k(t_1, \dots, t_k) \bar{P}_k(t_1, \dots, t_k)^T dt_1 \dots dt_k,$$

where

$$\bar{P}_1(t_1) = e^{At_1}B, \qquad \bar{P}_2(t_1, t_2) = \sum_{k=1}^m e^{At_1} N_1^k e^{At_2}B,$$
$$\bar{P}_3(t_1, t_2, t_3) = e^{At_1} H_2 e^{At_2} B \otimes e^{At_3}B, \qquad \dots$$

are the kernels of the Volterra series expansion of the system output.



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are the kernels of the Volterra series expansion of the system output.

Theorem[B./GOYAL/PONTES DUFF 2018]The reachability Gramian P of a PC system solves the polynomial Lyapunov equation
$$AP + PA^T + BB^T + \sum_{j=2}^{n_p} H_j (\otimes^j P) H_j^T + \sum_{j=2}^{n_p} \sum_{k=1}^m N_j^k (\otimes^j P) \left(N_j^k\right)^T = 0.$$



The Observability Gramian is defined as follows:

• First, we write the adjoint system as

[FUJIMOTO ET AL. 2002]

$$\begin{split} \dot{x}(t) &= Ax(t) + \sum_{j=2}^{n_p} H_j x_j^{\otimes}(t) + \sum_{j=1}^{n_p} \sum_{k=1}^m N_j^k x_j^{\otimes}(t) u_k(t) + Bu(t), \\ \dot{x}_d(t) &= -A^T x_d(t) - \sum_{j=2}^{n_p} H_j^{(2)} x_{d,j}^{\otimes}(t) - \sum_{j=1}^{n_p} \sum_{k=1}^m \left(N_j^{k,(2)} \right) x_{d,j}^{\otimes}(t) u_{d,k}(t) - C^T u_d(t), \quad x_d(\infty) = 0, \\ y_d(t) &= B^T x_d(t). \end{split}$$



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• Then, by taking the kernel of Volterra series, one has

Theorem

[B./GOYAL/PONTES DUFF 2018]

Let P be the reachability Gramian. Then, the observability Gramian Q of the PC system solves the polynomial Lyapunov equation

$$A^{T}Q + QA + C^{T}C + \sum_{j=2}^{n_{p}} H_{j}^{(2)} \left(\otimes^{j-1}P \otimes Q \right) \left(H_{j}^{(2)} \right)^{T} + \sum_{j=2}^{n_{p}} \sum_{k=1}^{m} N_{j}^{k,(2)} \left(\otimes^{j-1}P \otimes Q \right) \left(N_{j}^{k,(2)} \right)^{T} = 0.$$



- Polynomial Lyapunov equations are very expensive to solve, efficient algorithms have not yet been developed.
- We thus propose truncated Gramians that only involve a finite number of kernels and can be computed using the methods in MORLAB or M-M.E.S.S.:

$$P_{\mathcal{T}} = \sum_{k=1}^{n_p+1} \int_0^\infty \cdots \int_0^\infty \bar{P}_k(t_1, \dots, t_k) \bar{P}_k(t_1, \dots, t_k)^T dt_1 \dots dt_k.$$

Truncated Gramians

The truncated reachability Gramian solves

$$\begin{split} AP_{\mathcal{T}} + P_{\mathcal{T}}A^T + BB^T + \sum_{j=2}^{n_p} H_j \otimes^j P_l H_j^T + \sum_{j=2}^{n_p} \sum_{k=1}^m N_j^k \otimes^j P_l \left(N_j^k\right)^T = 0. \\ e AP_l + P_l A^T + BB^T = 0 \end{split}$$

• Advantage: Only need to solve a finite number of (linear) Lyapunov equations.

where



$$\epsilon v_t(x,t) = \epsilon^2 v_{xx}(x,t) + f(v(x,t)) - w(x,t) + q, w_t(x,t) = hv(x,t) - \gamma w(x,t) + q,$$

with a nonlinear function

$$f(v(x,t)) = v(v - 0.1)(1 - v),$$

 $\epsilon=0.015, \ h=0.5, \ \gamma=2, \ q=0.05, \ L=0.2,$ and boundary conditions:

$$v_x(0,t) = i_0(t), \quad v_x(L,t) = 0, \quad t \ge 0,$$

- Spatial discretization leads to PC system with cubic nonlinearity of order $n_{pc} = 600$.
- Lifting: $z := v^2 \Rightarrow f(v, z) = -vz + 1.1z 0.1v, \quad z_t = 2vv_t = \dots \rightsquigarrow$
- lifted quadratic-bilinear (QB) system of order $n_{qb} = 900$.
- Outputs of interest v(0,t), w(0,t) are the responses at the left boundary.
- We compare balanced truncation for PC and QB systems.





• Decay of singular values for PC systems is faster \Rightarrow smaller reduced-order model!





- Original PC system of order 600. Original QB system of order 900.
- $\bullet\,$ Reduced PC system of order $10.\,$ Reduced QB system of order $10.\,$





- Original PC system of order 600. Original QB system of order 900.
- Reduced PC system of order 10. Reduced QB system of order 30.





- Original PC system of order 600. Original QB system of order 900.
- Reduced PC system of order 10. Reduced QB system of order 43.



BT for linear systems:

- Method of choice for model order reduction in optimal and feedback control.
- Computational efficiency enhanced through advanced techniques from numerical linear algebra so that problems of industrial scale can be reduced.
- Numerous technical details necessary for use in digital and virtual twins.
- Robust and efficient implementations available in numerous software packages like SLICOT, M.E.S.S., MORLAB, pyMOR, ...



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BT for nonlinear systems:

- BT extended to bilinear, quadratic-bilinear, and polynomial systems.
- Not in this talk: local Lyapunov stability is preserved.
- As of yet, only weak motivation by local bounds of energy functionals.
- No error bounds in terms of "Hankel" singular values.
- Computationally efficient (as compared to nonlinear balancing), and input independent.
- To do:
 - improve efficiency of Lyapunov solvers with many right-hand sides further;
 - error bound;
 - conditions for existence of new PC Gramians;
 - extension to descriptor systems.



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Structure preserving model order reduction of large sparse second-order index-1 systems and application to a mechatronics model



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Approximate Balanced Truncation for Polynomial Control Systems. Coming soon(er or later).


Commercial: 3-Volume Handbook "Model Order Reduction"



- Edited by Peter Benner, Stefano Grivet-Talocia, Alfio Quarteroni, Gianluigi Rozza, Wil Schilders, and Luís Miguel Silveira,
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